

# POWER ELECTRONICS

## Key Concepts

**P. T. Krein**

**Department of Electrical and Computer Engineering  
University of Illinois at Urbana-Champaign**

© 2011, 2015 Philip T. Krein. This work is licensed under a  
Creative Commons Attribution-NonCommercial-ShareAlike 4.0 International License except certain photos as labeled.



## Notation Points

- Upper case for dc quantities.  $V_{in}$
- Lower case for ac quantities.  $v(t)$
- Angle brackets  $\langle \rangle$  for averages.

$$\frac{1}{T} \int_0^T f(t) dt = \langle f(t) \rangle = F$$

## Notation Points

- Time-varying average (moving average):

$$\frac{1}{T} \int_{t-T}^t f(\tau) d\tau = \bar{f}(t) = F(t)$$

$$f(t) = F(t) + \tilde{f}(t)$$

$F(t)$ : Moving average

$\tilde{f}(t)$ : Time variation

## Time-Varying Average (Moving average)

$$\frac{1}{T} \int_{t-T}^t f(\tau) dt = \bar{f}(\tau)$$



## Notation Points

RMS (Root Mean Square):

$$F_{RMS} = \sqrt{\frac{1}{T} \int_0^T f^2(t) dt}$$

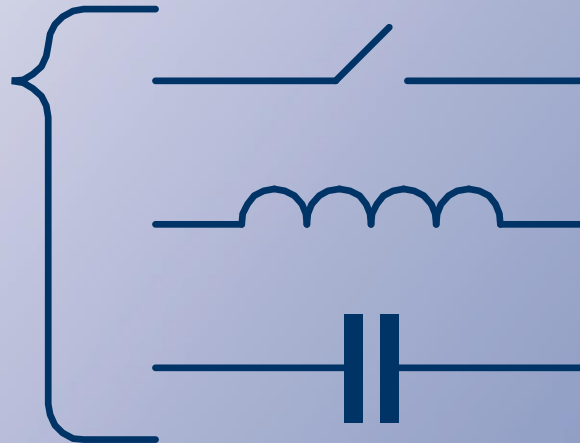
If  $f(t) = V_0 \cos(\omega t)$  then  $F_{RMS} = \frac{V_0}{\sqrt{2}}$

## Power Electronic System

Electrical source.

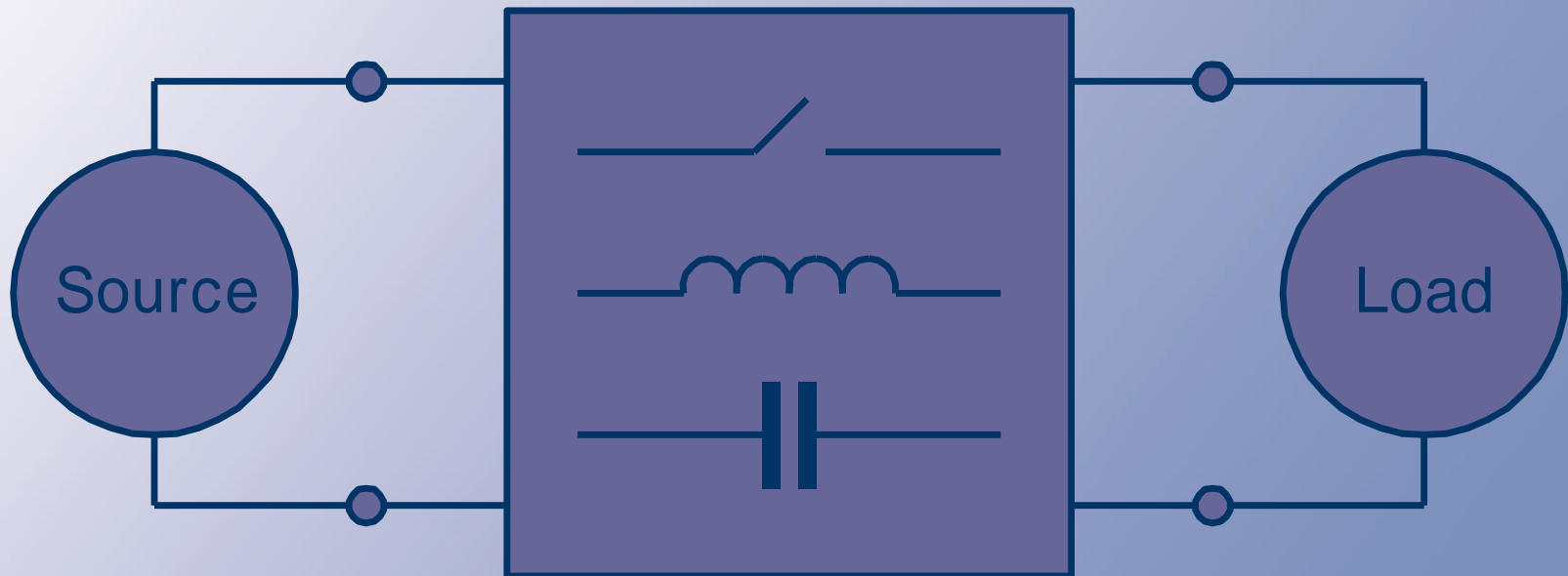
Conversion.

Load.



## Power Electronic System

### Conversion components



## The Switch Matrix

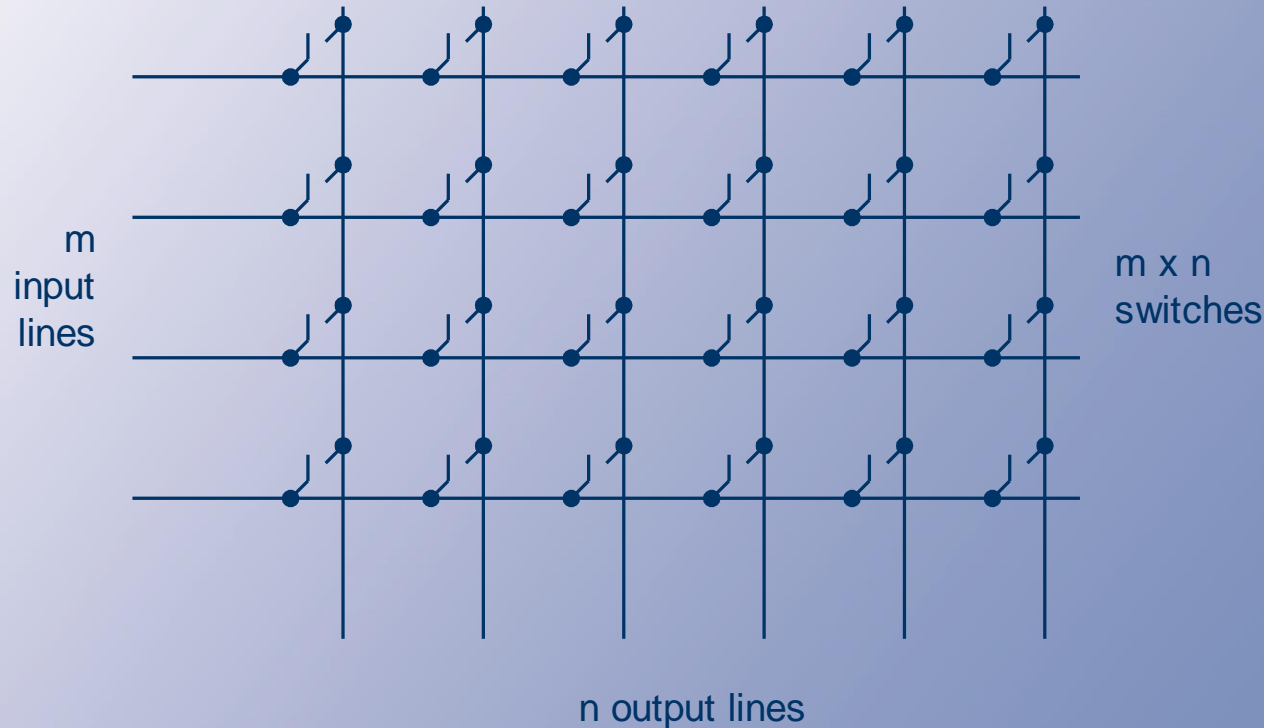
If a converter has  $m$  input lines and  $n$  output lines, an  $m \times n$  matrix allows all possible interconnections.



## Conversion

- Consider switches alone (no storage elements). The most complicated arrangement possible is:

SWITCH MATRIX

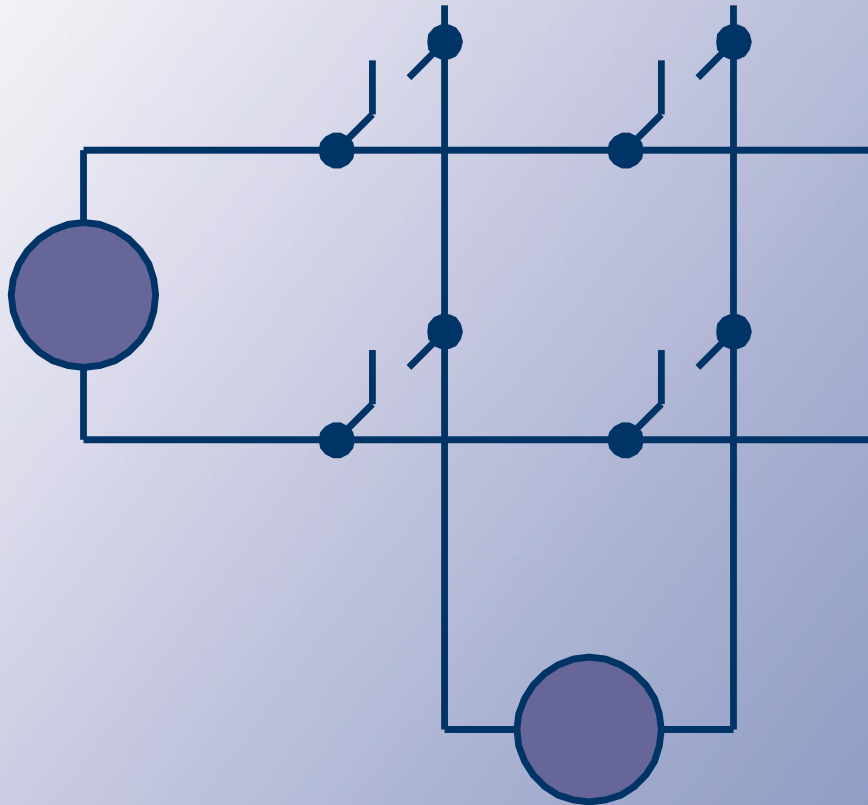


## Conversion

- Polyphase case:
  - Three inputs, three outputs
- High-voltage dc:
  - Up to 48 or more input lines
  - Perhaps 6 output lines
- Typical case:
  - 2 x 2 matrix

## Conversion

- Typical case with only 4 switches:



$2 \times 2 =$   
4 switches

## Power Electronics Focus

1) Build a switch matrix.

**Hardware.**

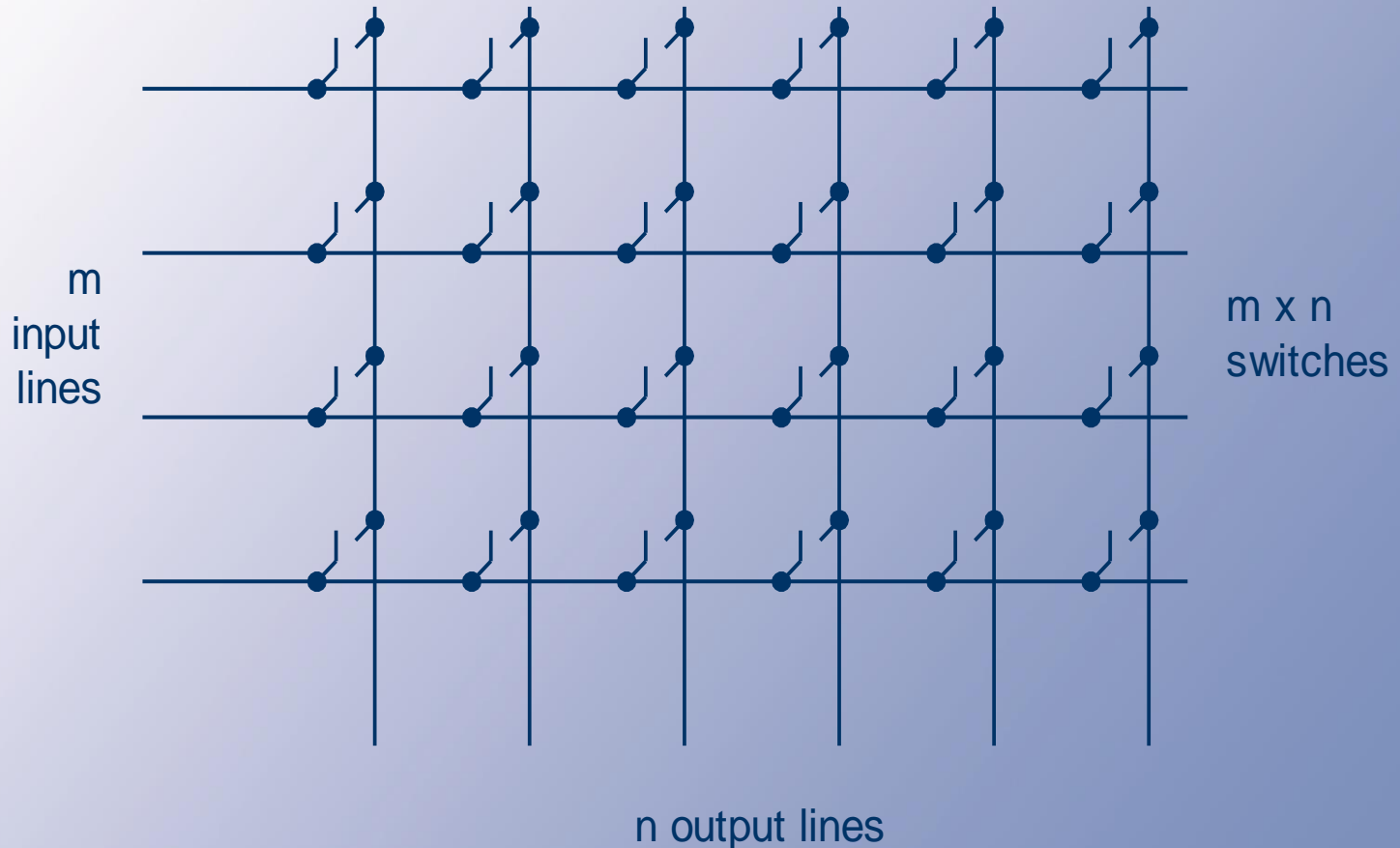
2) Determine how to operate matrix to get a desired result.

**Software.**

3) Use storage elements to interface with in/out.

**Interface.**

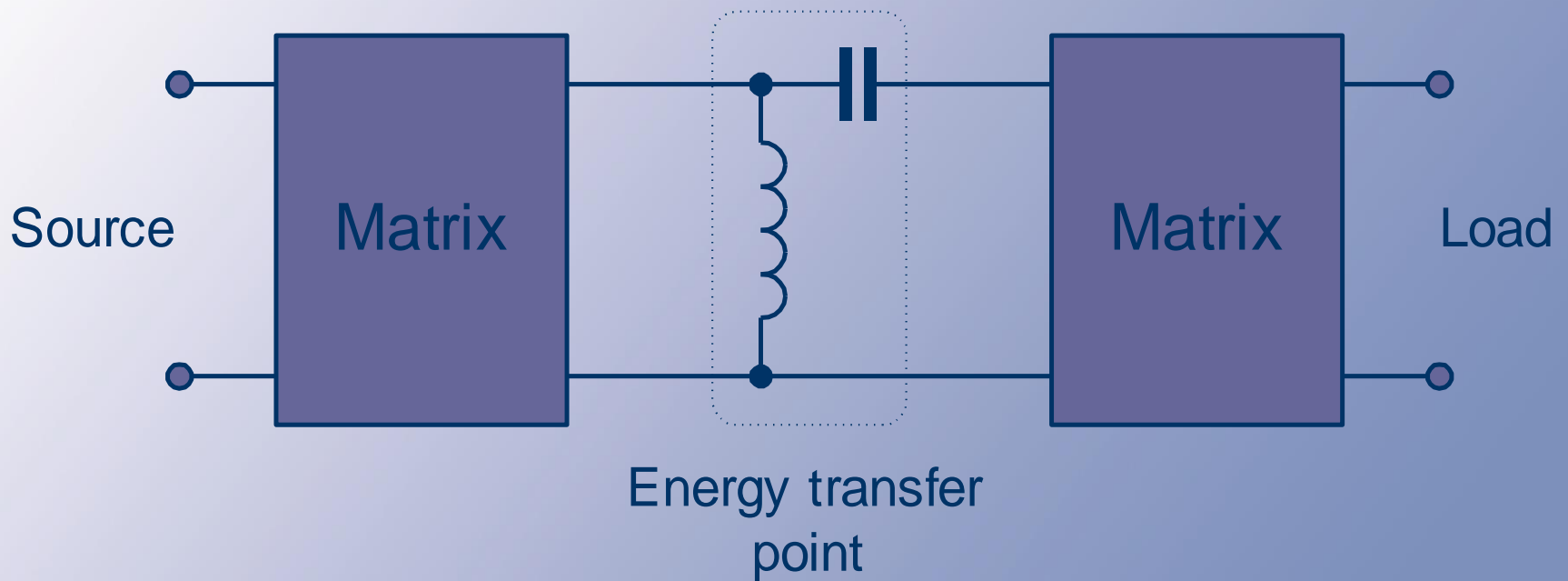
## Conversion



**Direct switch matrix (No storage)**

## Conversion

### Indirect switch matrix



## Methods

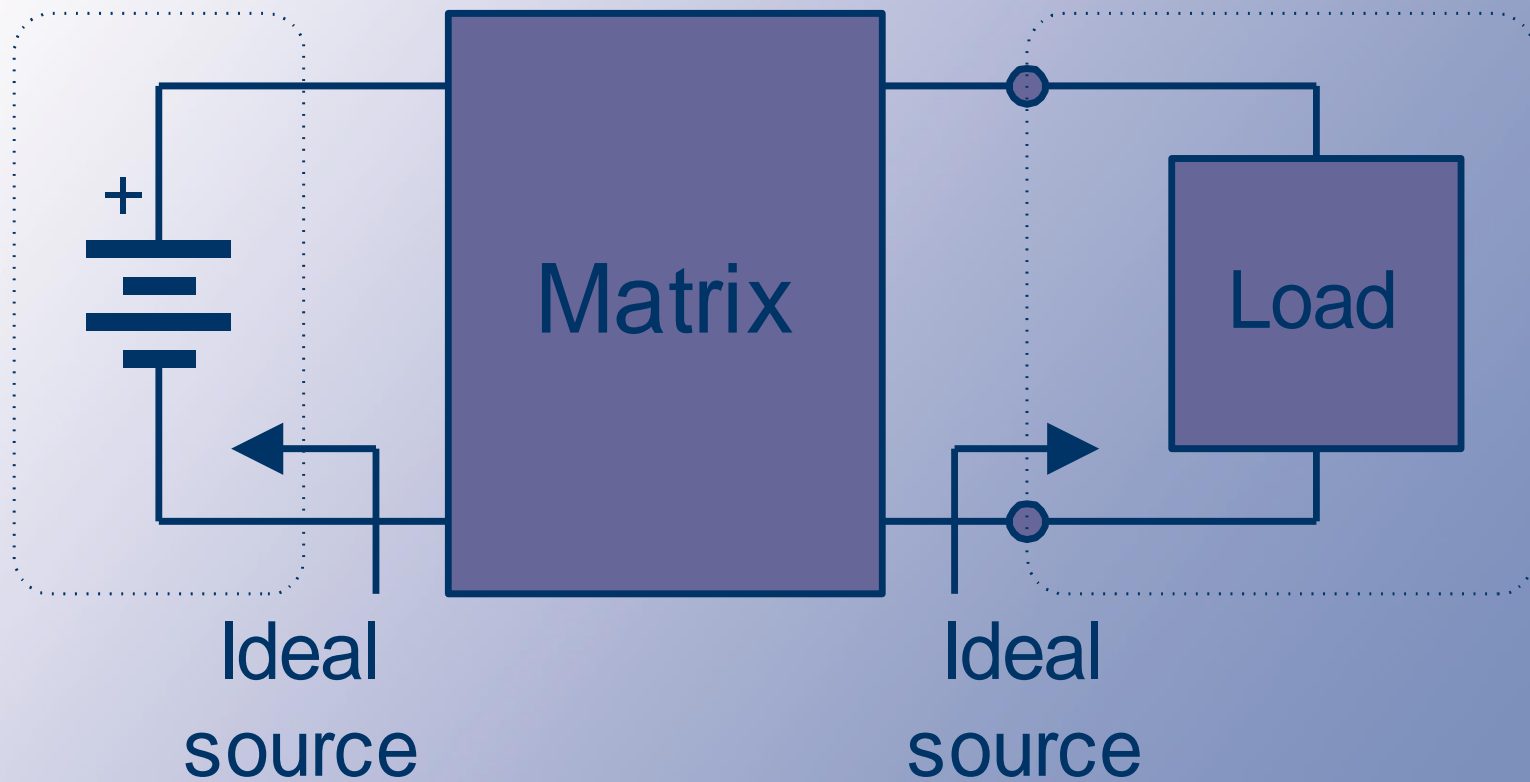
- The usual method for design and analysis of indirect switch matrix converters is to cascade two direct switch matrices and place the storage components between them.
- Cascaded converters are common:
  - Rectifier-inverter sets for motor drives
  - Rectifier-dc sets for power supplies
  - Dc-dc converter cascades for flexibility

## Source Conversion

- Energy conversion is not a generic process.
- User expectations:
  - **Voltage source.**
- Generalization:
- Electrical input: (Ideal) **Source.**
- Electrical load: User wants ideal source.

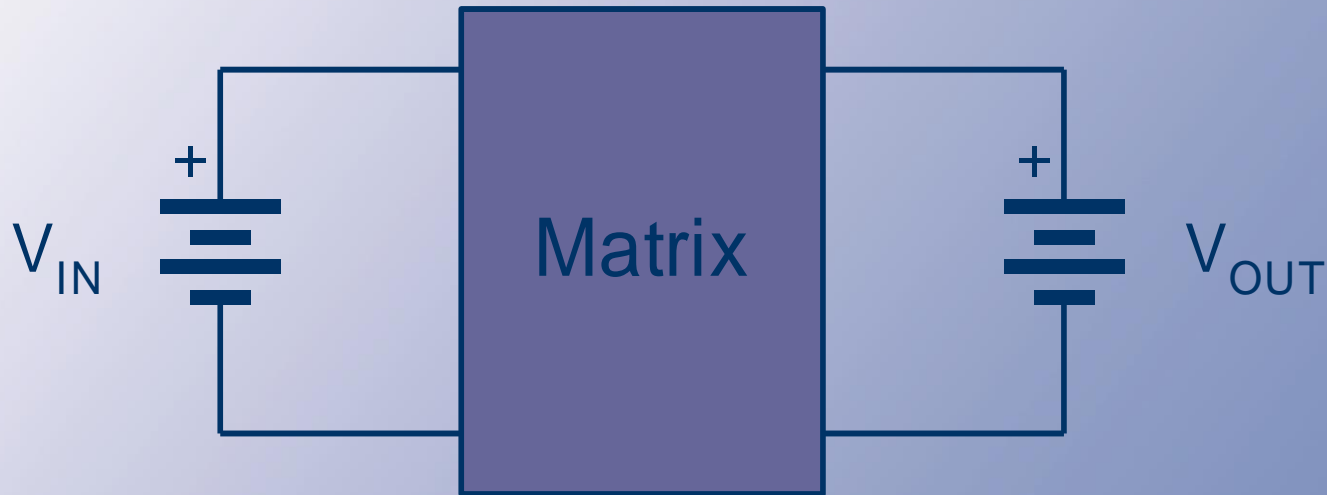


## Source Conversion



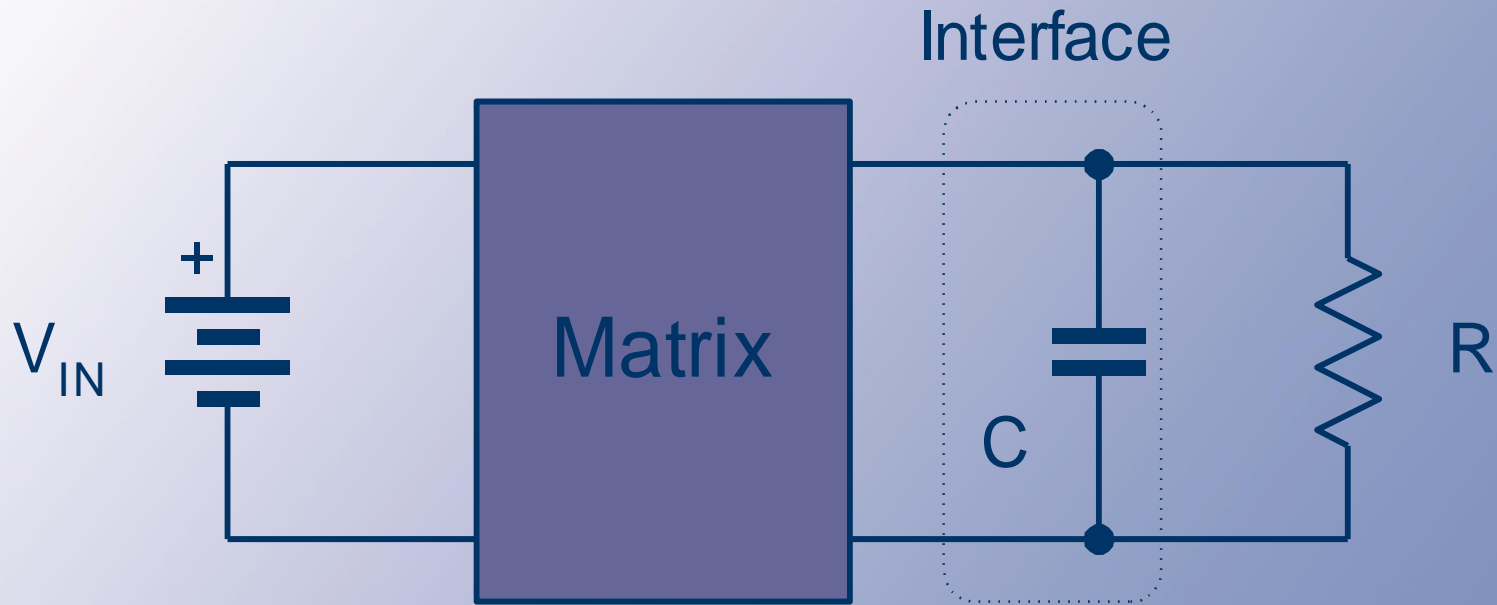
## Source Conversion

Implications: We can analyze the circuit this way.



Power Electronic Circuit: Exchange energy among ideal sources.

## Source Conversion

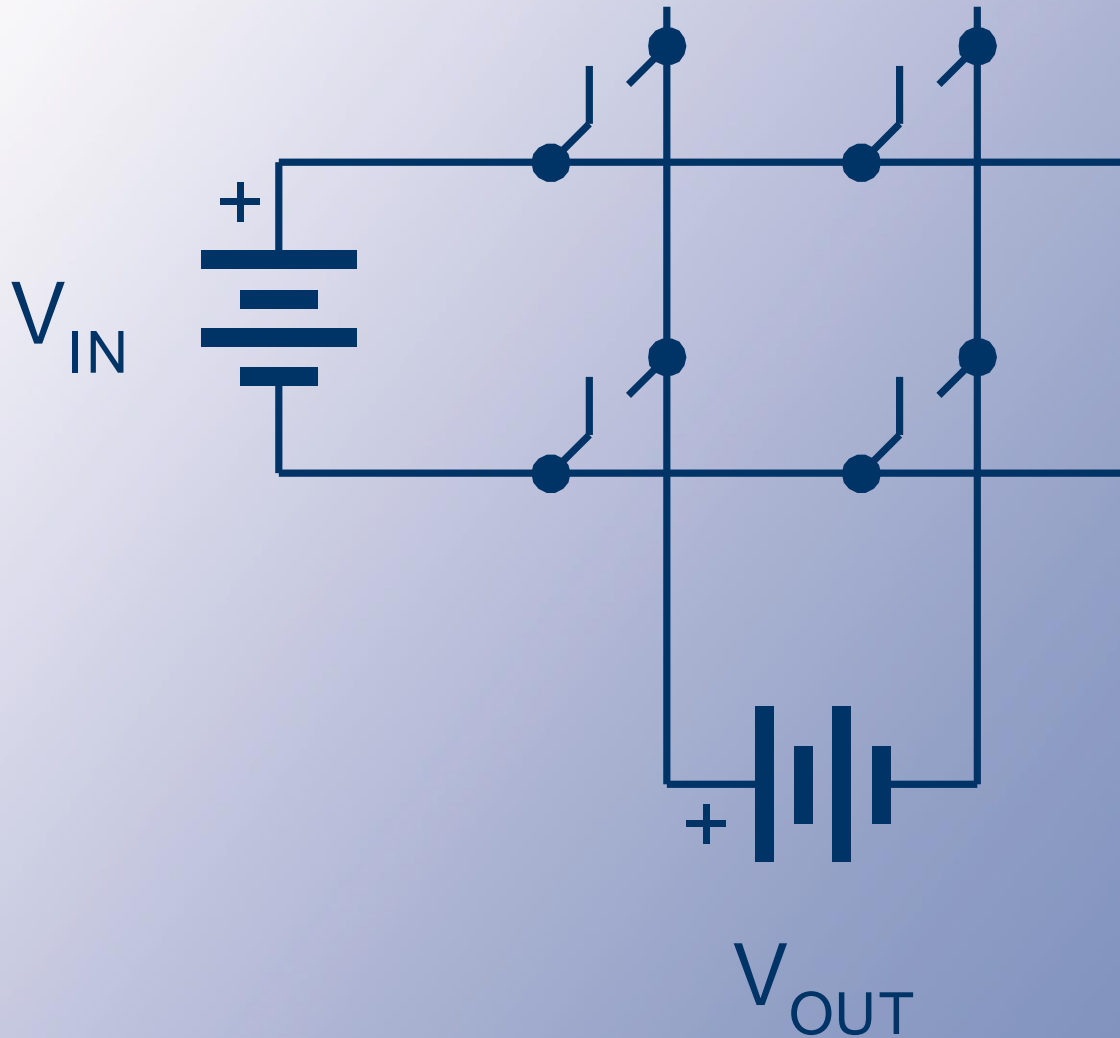


Large  $C \Rightarrow$  Little change in  $V_c (= V_{out})$ .

## Source Conversion

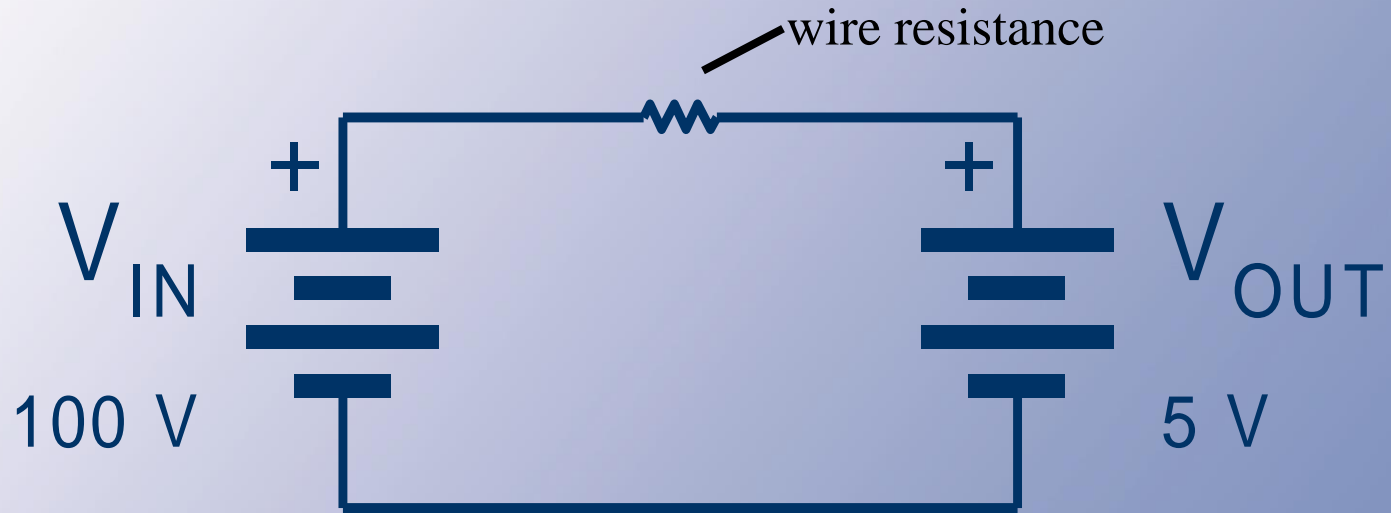
- The following slide shows a logical source conversion approach.
- The input and output are ideal dc voltages, and a switch matrix sits between them as a direct converter.
- Check the configurations.

## Source Conversion



## Source Conversion

One configuration would be:



Circuit laws:

$$\text{KVL} : \sum V_{\text{loop}} = 0$$

$$\text{KCL} : \sum I_{\text{node}} = 0$$

## Circuit Laws

- There is a problem here!
- It would seem that the sum of voltages around the loop is nonzero!
- The reality is that wires and real devices have some (small) resistance.
- A large current will flow – and we hope will blow a fuse.

## Circuit Laws

- KVL problem: Cannot interconnect unlike voltages.
- Trouble: switches do not “know” KVL.
- Power converter: Can **attempt** a violation.
- But a violation will not really occur – only a problem (or a fire).





## The Reality of KVL

- We see that KVL has a concrete meaning in power electronics.
- The Law becomes: *Do not interconnect unlike voltage sources.*
- There is a real and often costly penalty for a violation.
- For design, we can think in terms like “Do not even *try* to violate KVL.”

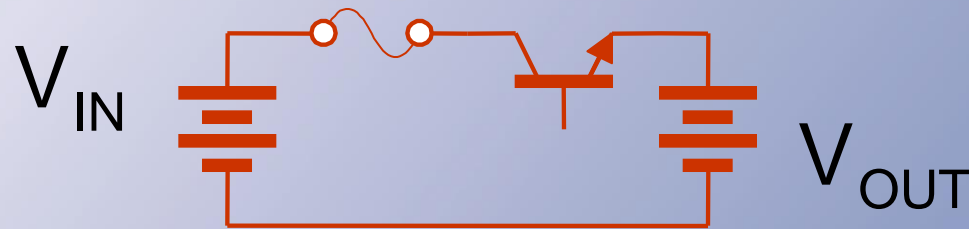
## Reality of Circuit Laws

$$\text{KVL: } \sum v_{\text{loop}} = 0$$

Implication: unequal voltage sources cannot be interconnected.

In power electronics (not in other fields), we can build a circuit that tries to “violate” KVL. Attempts to connect unlike voltages yield extreme currents and failures.

## Reality of Circuit Laws



- In the top circuit, the switch must be OFF so that KVL is not violated.
- In the bottom circuit, unfortunately, the transistor will probably fail before the fuse does anything.

## Reality of Circuit Laws

$$\text{KCL: } \sum i_{\text{node}} = 0$$

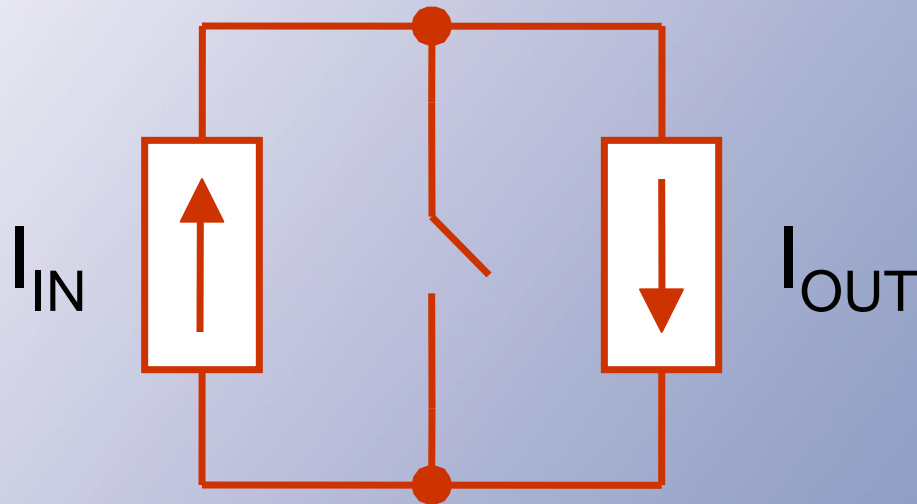
- The implication is that unequal current sources cannot be interconnected.
- It is possible to build a circuit that tries to “violate” KCL.

Example: **Inductor**  
carrying current.  
**Disconnect it abruptly!**

Current flow continues

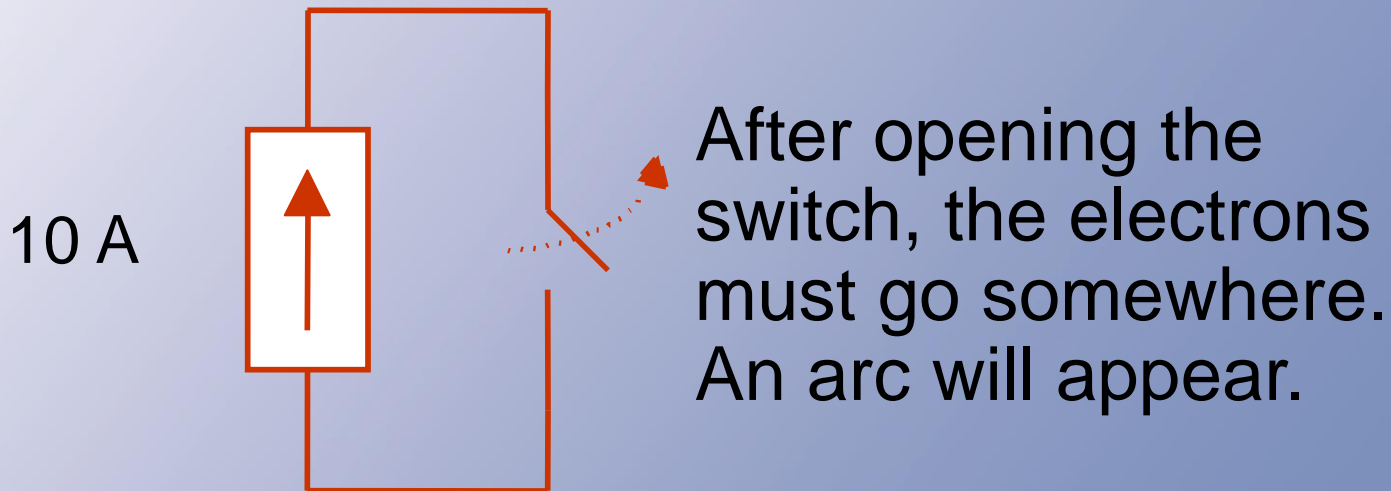
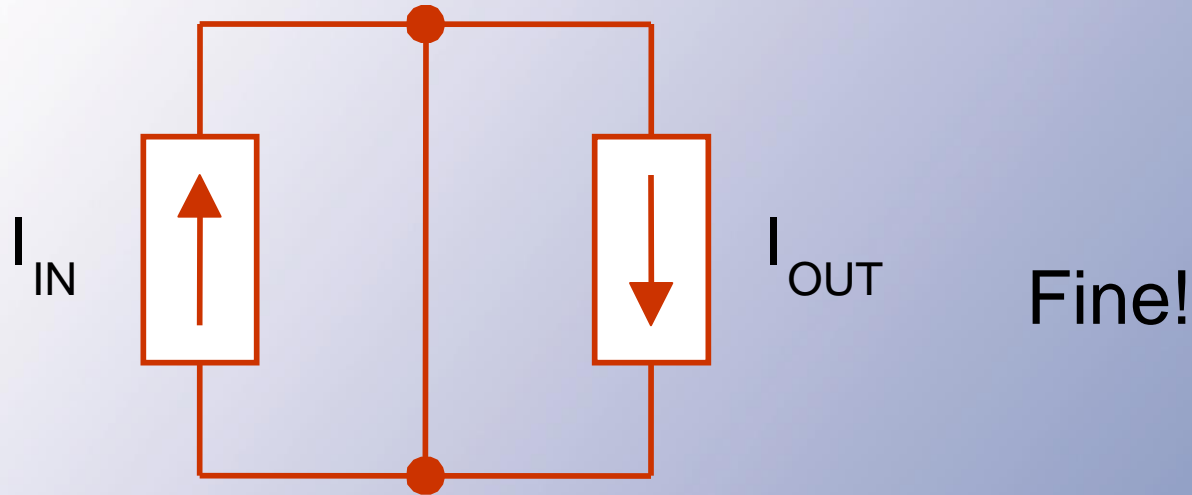


## Reality of Circuit Laws

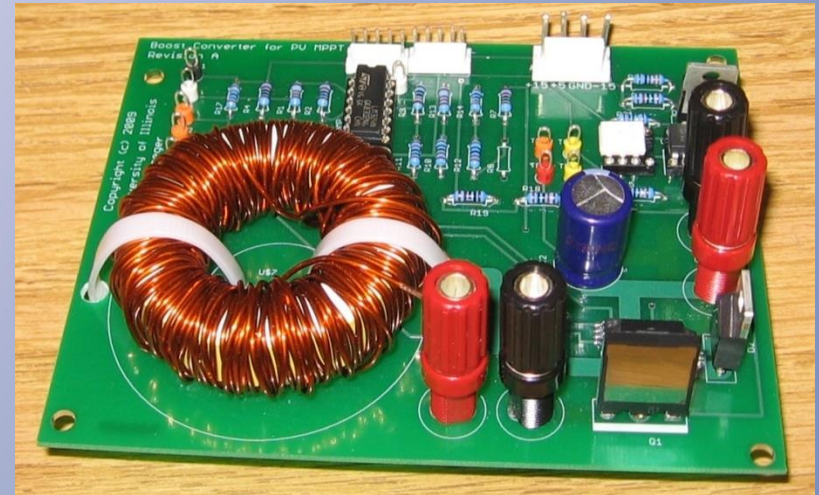
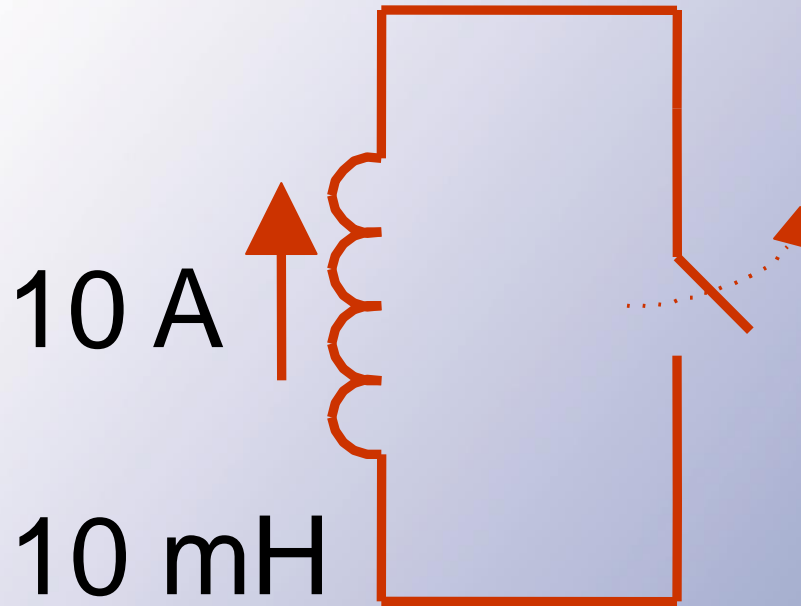


The switch must be ON, so that KCL is not violated.

## Reality of Circuit Laws

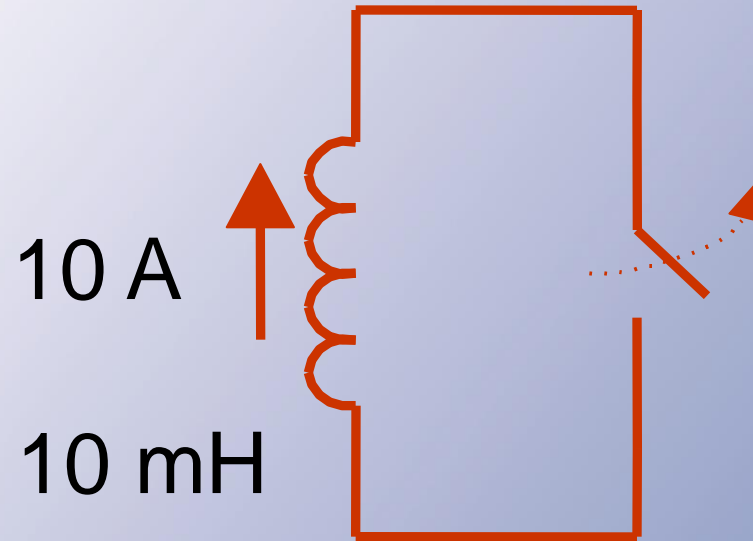


## One-Port Model



$v = L \, di/dt$ . If current drops in  $1 \, \mu\text{s}$ , we have:

## One-Port Model

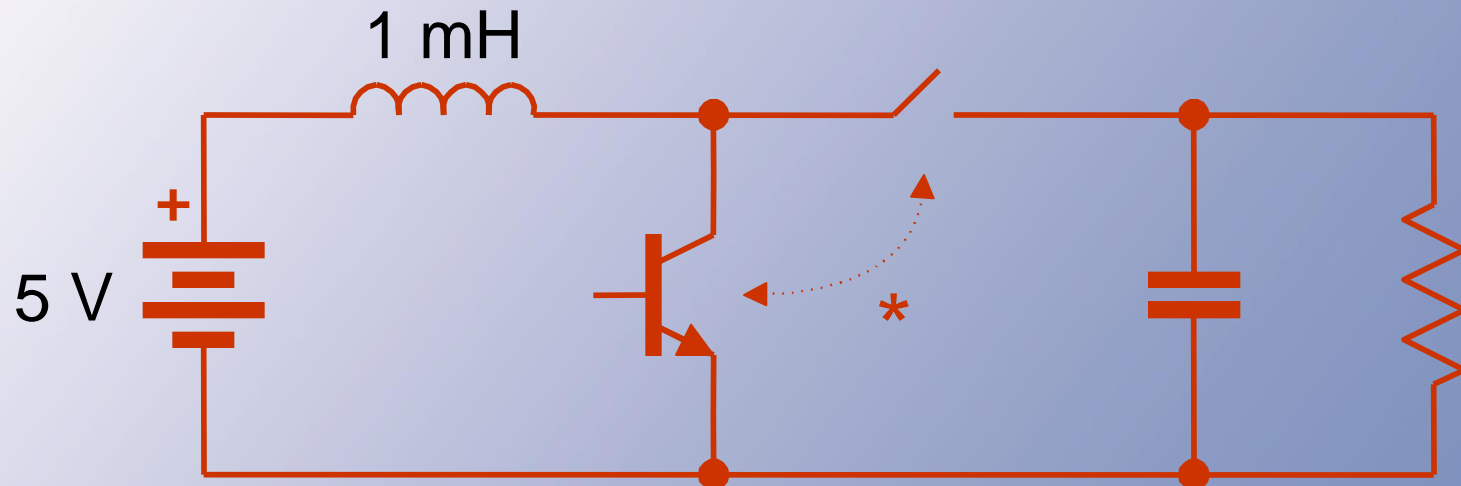


$$\begin{aligned}v_L &= L \, di/dt \\ &= (0.01 \text{ H}) \frac{-10 \text{ A}}{1 \mu\text{s}} \\ &\approx -100,000 \text{ V}\end{aligned}$$



## Reality of Circuit Laws

KCL: Must provide a current path for any current source.



\* Switching must be coordinated correctly, so as not to remove the path for  $I_L$ .

## Reality of Circuit Laws

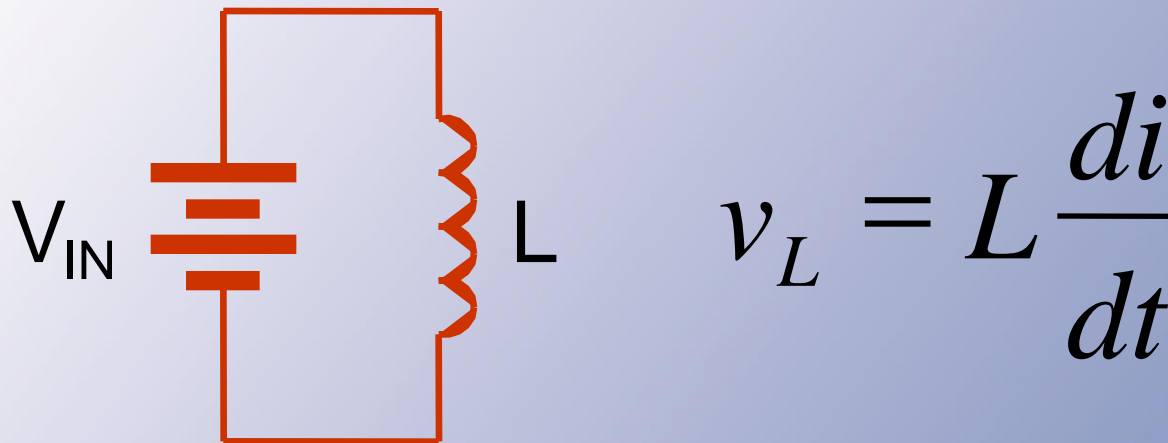
- Attempts to violate KCL can generate extreme voltages, as current tries to maintain its flow.
- It is hard to protect against this – fuses do not help.
- KVL “violations” are reasonable easy to avoid.
- KCL is more problematic in practice.



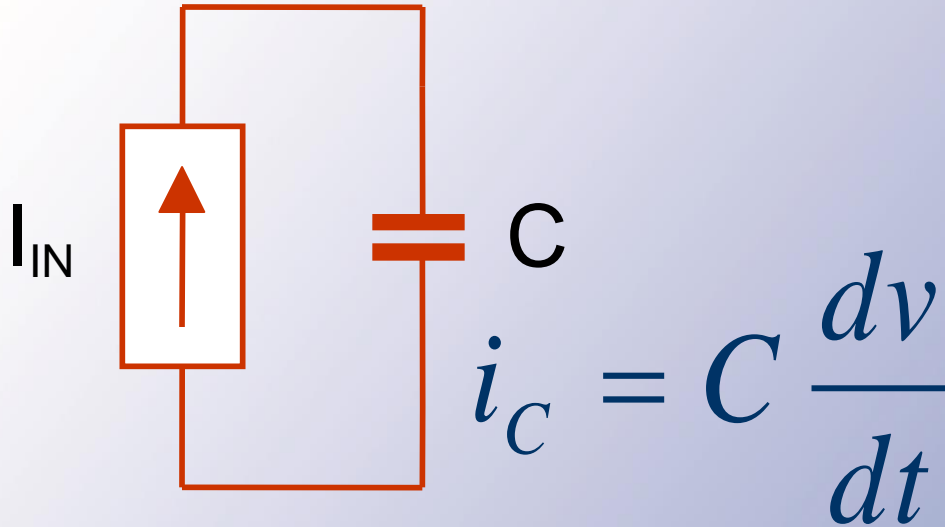
## Implications for Storage

- If a fixed voltage is applied to an inductor, current rises without limit.
- This is like a short circuit, although it is OK for a short time.
- If a fixed current is applied to a capacitor, charge rises without limit.

## Implications for Storage



Okay in general, but must be time-limited, if we apply dc voltage to an inductor. In the ideal arrangement, there is no limitation on the current. *An inductor will not sustain dc voltage.*



Okay, but must be time-limited.  
Cannot apply dc current to a capacitor.  
Capacitor will not sustain dc current.

## Implications for Storage

- An inductor cannot sustain dc voltage over extended times.
- A capacitor cannot sustain dc current over extended times.

## Implications for Storage

Since  $v_L$  must have no dc and  $i_C$  must have no dc, it must be true that:

$$\langle v_L \rangle = 0, \quad \langle i_C \rangle = 0.$$

These are key to circuit analysis: an inductor carries no average voltage; a capacitor carries no average current.

$$\langle v_L \rangle = 0$$

$$\langle i_C \rangle = 0$$

→ KVL and KCL

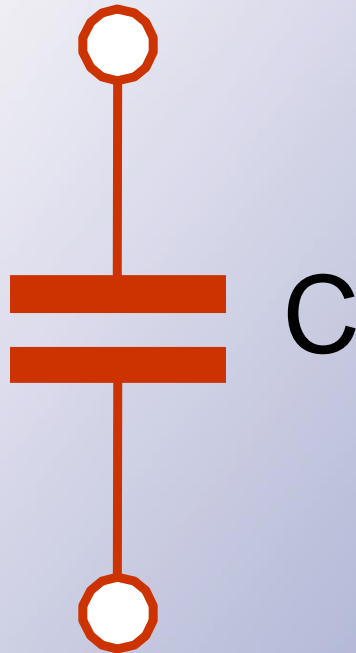
## Implications for Switching

- We want ideal sources (source conversion concept).
- We cannot use a switch matrix for direct connection of voltage sources or of current sources.





## Implications for Switching



$$i_C = C \frac{dv}{dt}$$

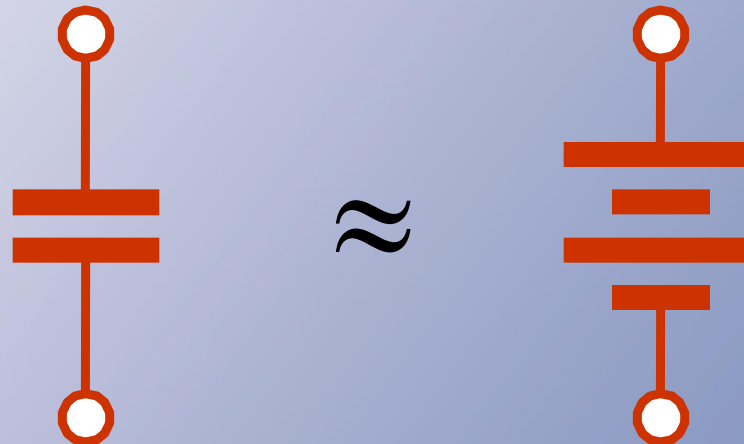
Over short times:

$$\frac{dv_C}{dt} = \frac{i_C}{C}$$

If  $C$  is large and  $i_C$  is bounded, then  $dv/dt$  can be as small as desired.

## Implications for Switching

The previous statement means that a large capacitor acts as a voltage source over short times.



## Implications for Switching



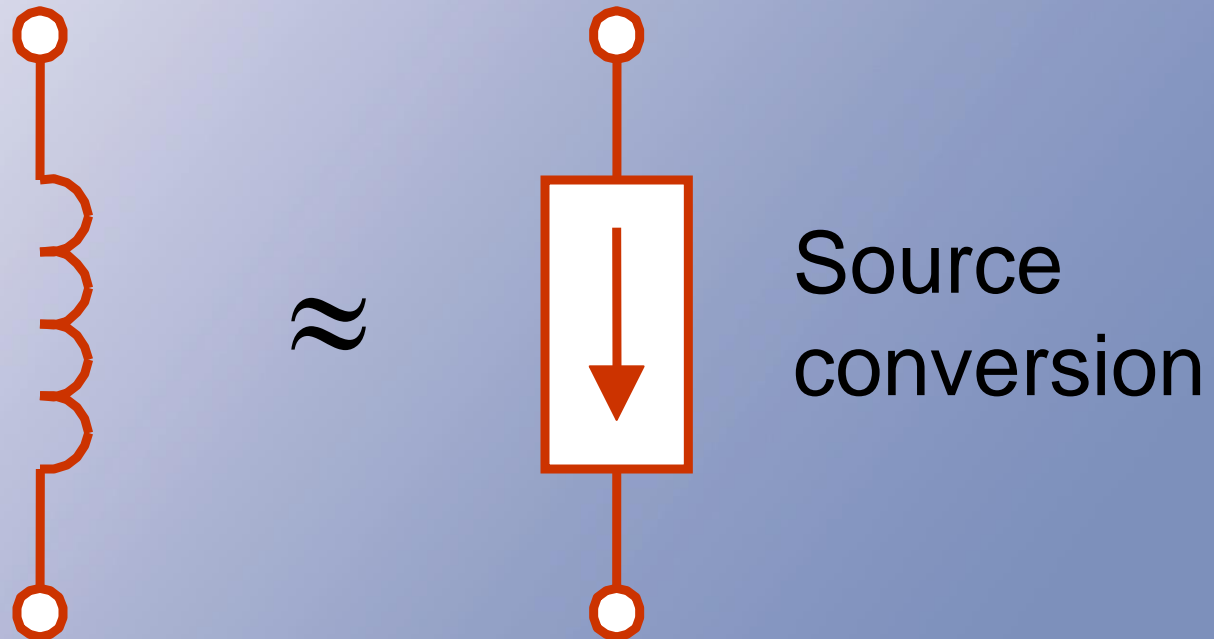
$$v_L = L \frac{di}{dt}$$

$$\frac{di_L}{dt} = \frac{v_L}{L}$$

If  $L$  is large and  $v_L$  is bounded, then  $di/dt$  can be made as small as desired.

## Implications for Switching

The previous statement means that a large inductor acts as a current source over short times.

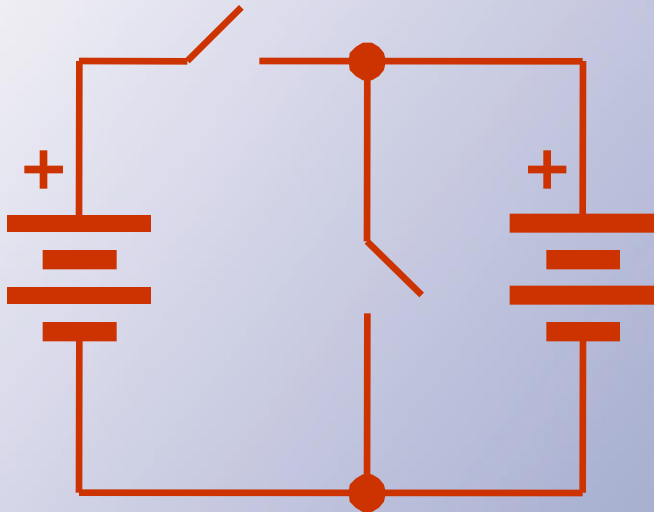


## Implications for Switching

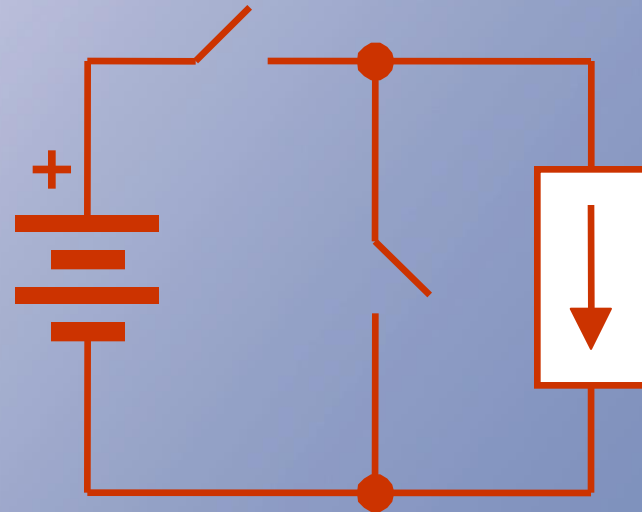
- Any useful converter must mix voltage and current sources.
- “Voltage converts to current,”
- “Current converts to voltage.”

## Implications for Switching

The user wants ideal sources but we can't just use either  $V$  or  $I$ . Must mix these.

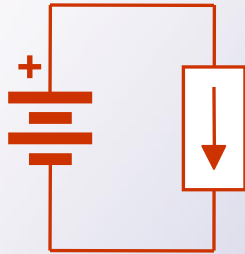


Problems

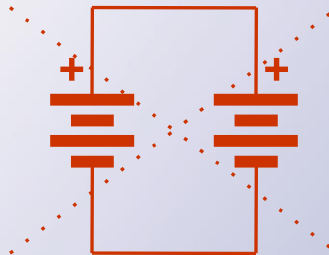


No problem

## Implications for Switching

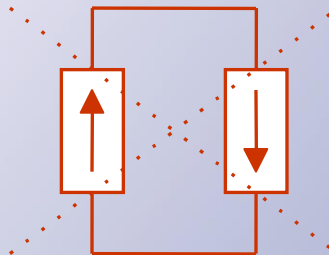


No problem



Problems

Example: Connecting two batteries in parallel, one ok, one discharged.



Problems

## Implications for Switching Source conversion

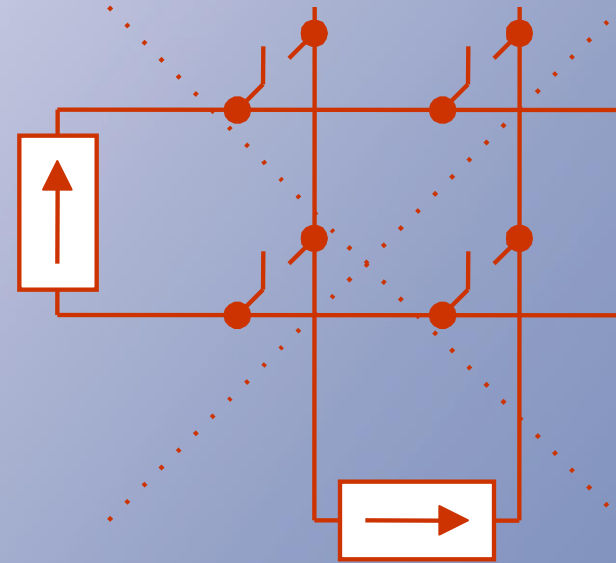
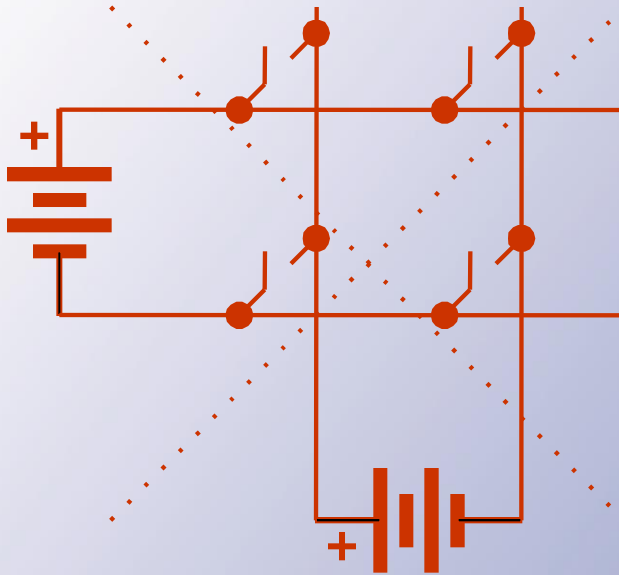
Voltage → Current

Current → Voltage



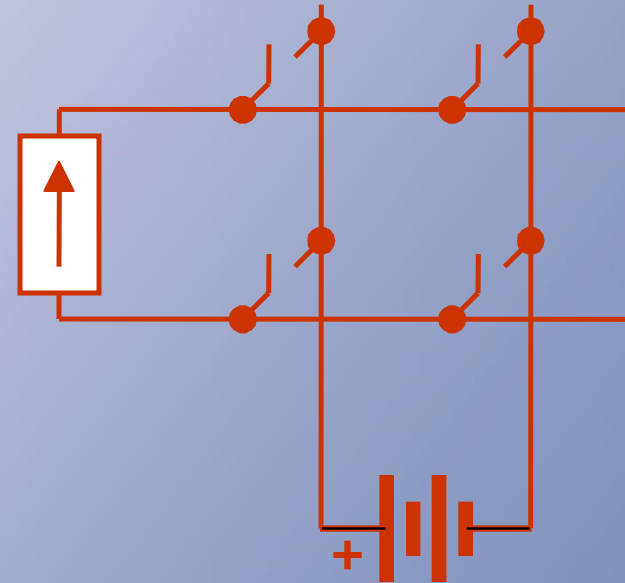
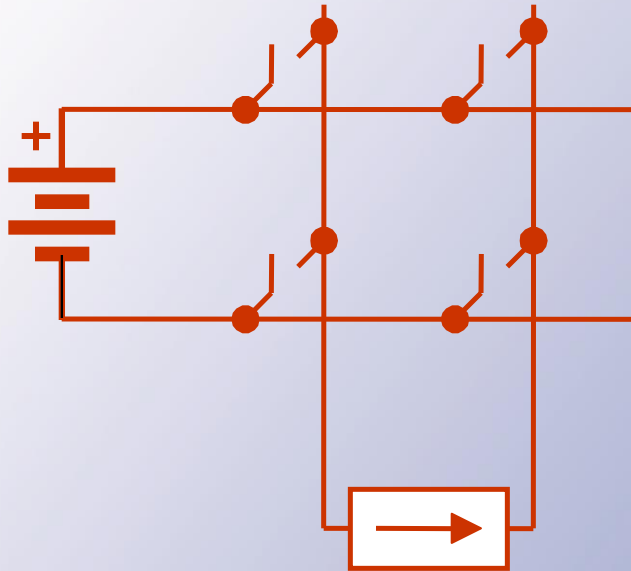


## Implications for Switching



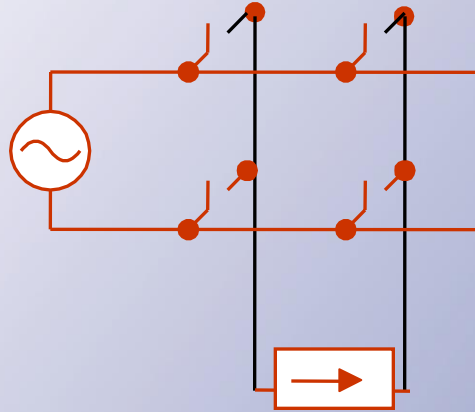
Because of KVL and KCL, neither of these can deliver useful energy.

## Implications for Switching

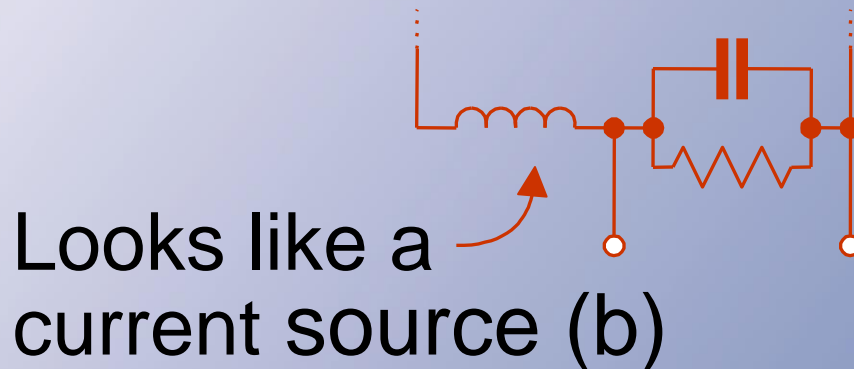


These are valid combinations.

## Implications for Switching Possible useful configuration (a)



We can achieve (a) with something like (b).



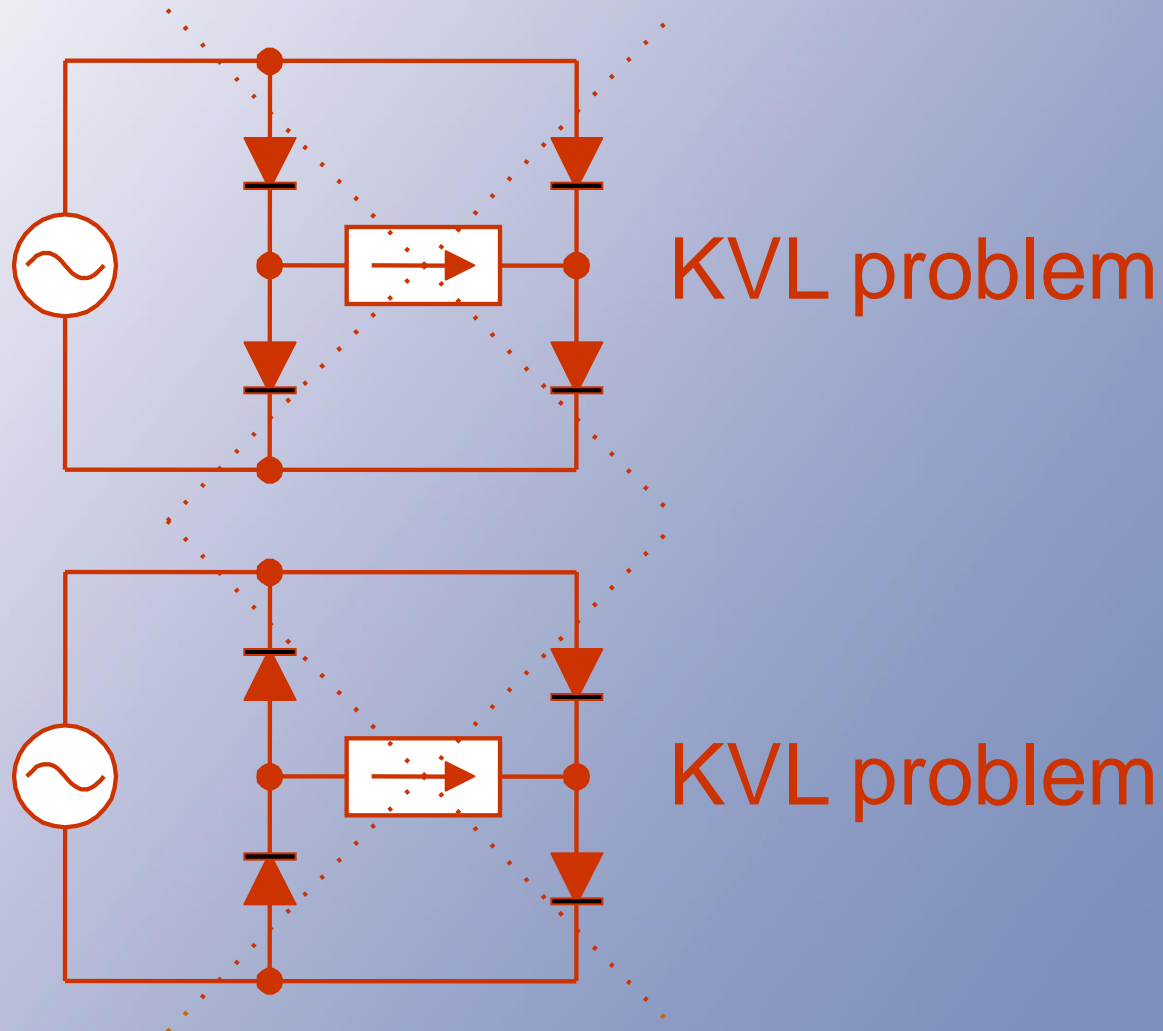
Looks like a current source (b)

## Diode Bridge Example

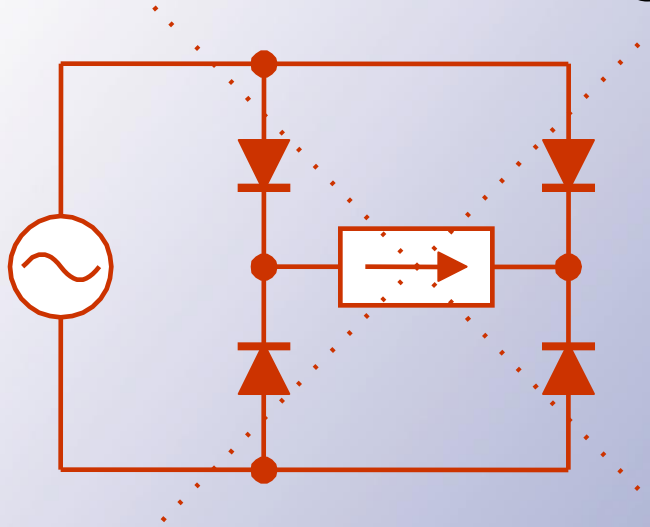
Of all possible connections, only one remains after KVL, KCL, and conversion requirements are met.



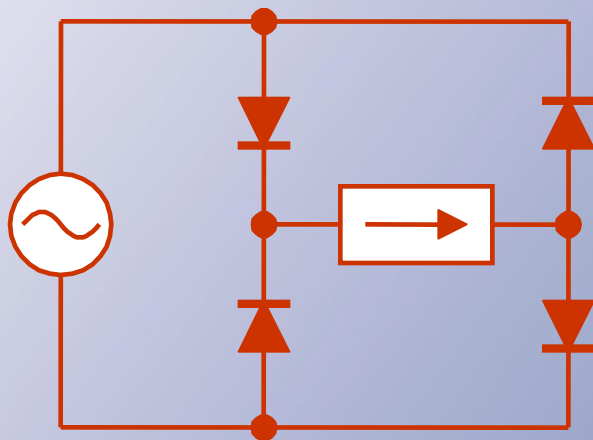
## Diode Bridge Example



## Diode Bridge Example



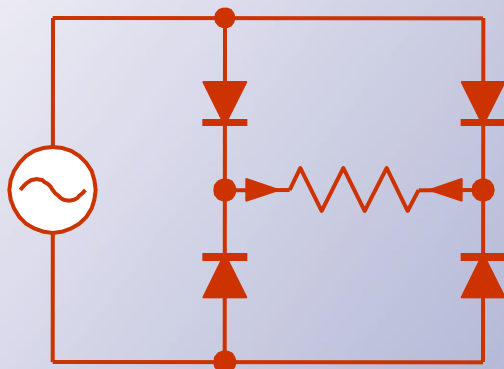
KCL problem



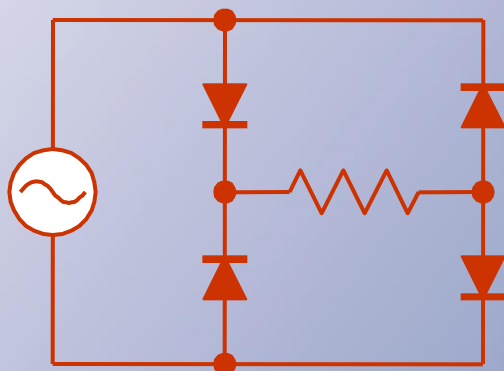
Diode Bridge.  
The only combination  
That does not violate  
KCL or KVL

## Diode Bridge Example

Can do the same with a resistor, rather than a current source. No KCL issues.



No KCL problem,  
but  $I_{\text{RESISTOR}} = 0$   
and therefore  
Power = 0



Diode Bridge.  
KVL and KCL are  
useful for us.

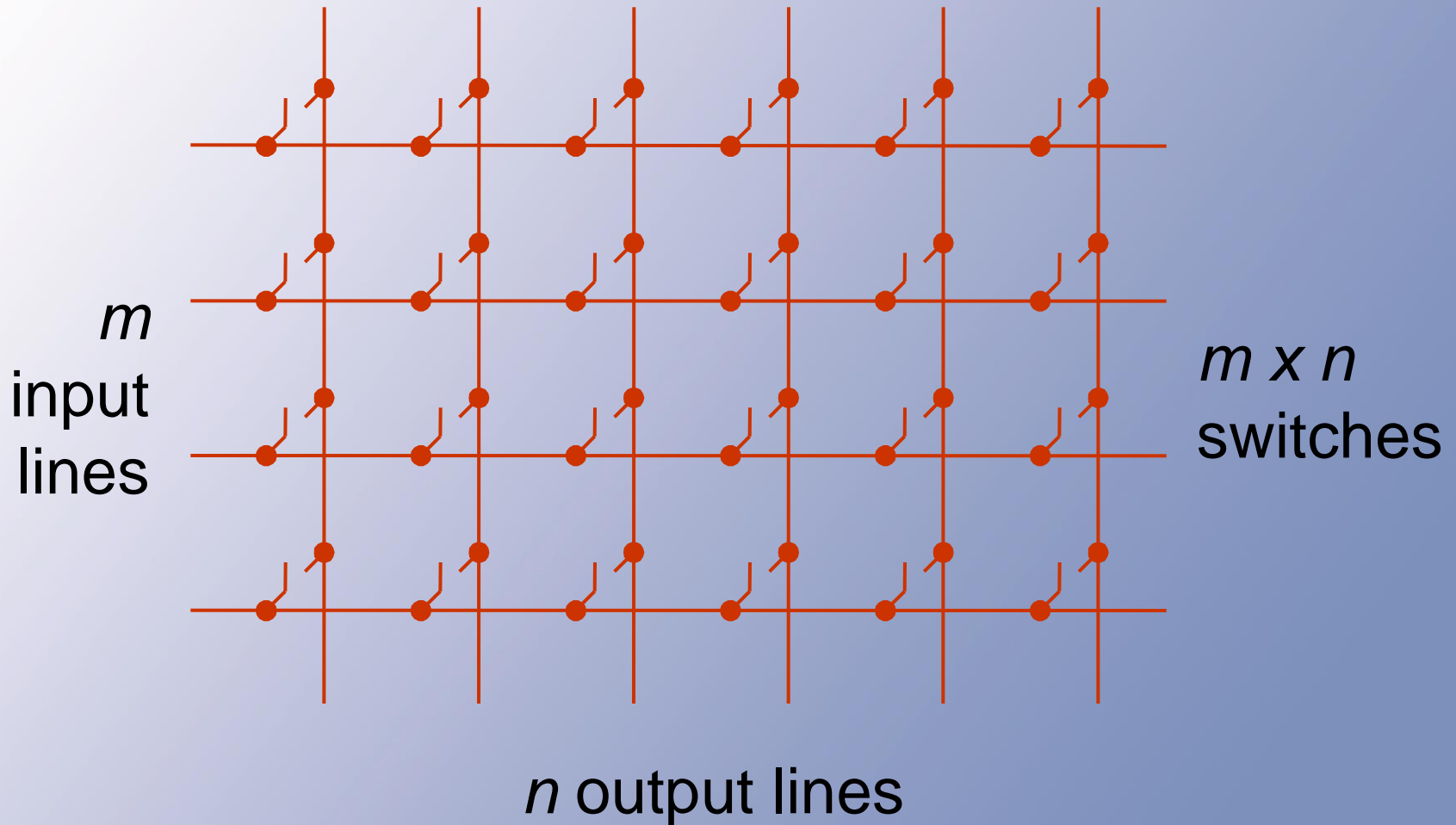
## Summary of Analysis Rules

1. Conservation of energy.
2. Source conversion.
3. KVL: avoid voltage source interconnection.
4. KCL: provide current paths.
5.  $\langle v_L \rangle = 0$
6.  $\langle i_C \rangle = 0$



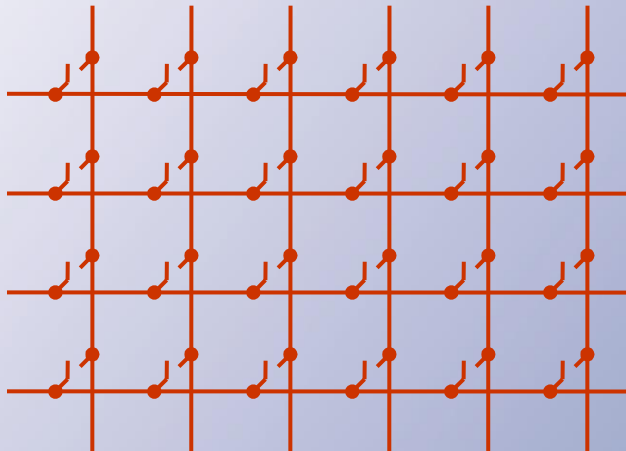


## Switching Functions



## Switching Functions

- The element at row  $i$  and column  $j$  represents switch  $ij$  and function  $q_{ij}(t)$ .
- The matrix  $Q$  has elements that correspond to each individual switch.



$$Q(t) = \begin{bmatrix} q & q & q & q & q \\ q & q & q & q & q \\ q & q & q & q & q \\ q & q & q & q & q \end{bmatrix}$$

$q = 1$  or  $0$  at any time

## Switching Functions

- We have a physical switch matrix with  $m$  rows and  $n$  columns.
- Each switch is either on or off.
- Define a *switching function*,  $q(t)$  as 1 when a device is on, 0 when off.



## Switching Functions

- Now, each physical switch is associated with a simple discrete function.
- Do not forget about time.
- We can define a *switch state matrix*,  $Q(t)$ .

## Switching Functions

- We can define our *software* problem in terms of choices of switching functions.
- We can find out the expected waveforms in many types of converters.

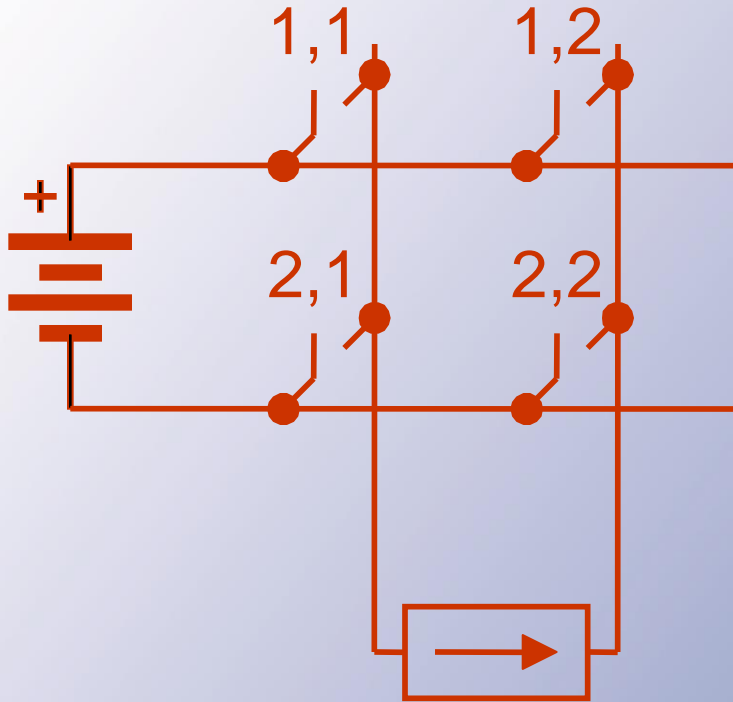
## Summary So Far

- KVL and KCL represent restrictions on what we can do.
- Switching functions make our actions easier to quantify and analyze.

## Switching Functions

- The functions support shorthand notation for KVL and KCL analysis.
- More important, they give mathematical expressions that represent converter action.

## Example



“Generic” DC-DC converter.

$$q_{11} = 0 \text{ or } 1$$

$$Q = \begin{bmatrix} q_{11} & q_{12} \\ q_{21} & q_{22} \end{bmatrix}$$



## Example

KVL says we **cannot have 1,1 and 2,1 ON or 1,2 and 2,2 ON, simultaneously.**

KVL: Not 1,1 + 2,1 ON together.

Not 1,2 + 2,2 ON together.

KCL: Not 1,1 + 2,1 OFF together.

Not 1,2 + 2,2 OFF together.

## Example

Compact description of the restrictions:

- KVL:  $q_{11} + q_{21} \leq 1$

$$q_{12} + q_{22} \leq 1$$

- KCL:  $q_{11} + q_{21} \geq 1$

$$q_{12} + q_{22} \geq 1$$

- Both:

$$q_{11} + q_{21} = 1$$

$$q_{12} + q_{22} = 1$$

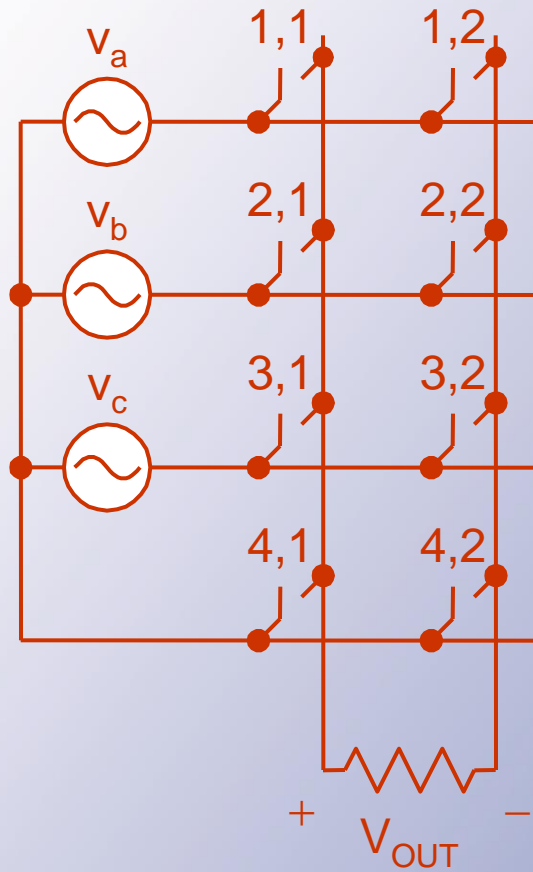
## Example

The last two expressions can be written together as:

$$\sum_{i=1}^N q_{ij} = 1$$

“One and only one switch on at a time, for each column.”

## Example 2



Three AC voltage sources connected through a switch matrix to a load. The sources share a reference point.

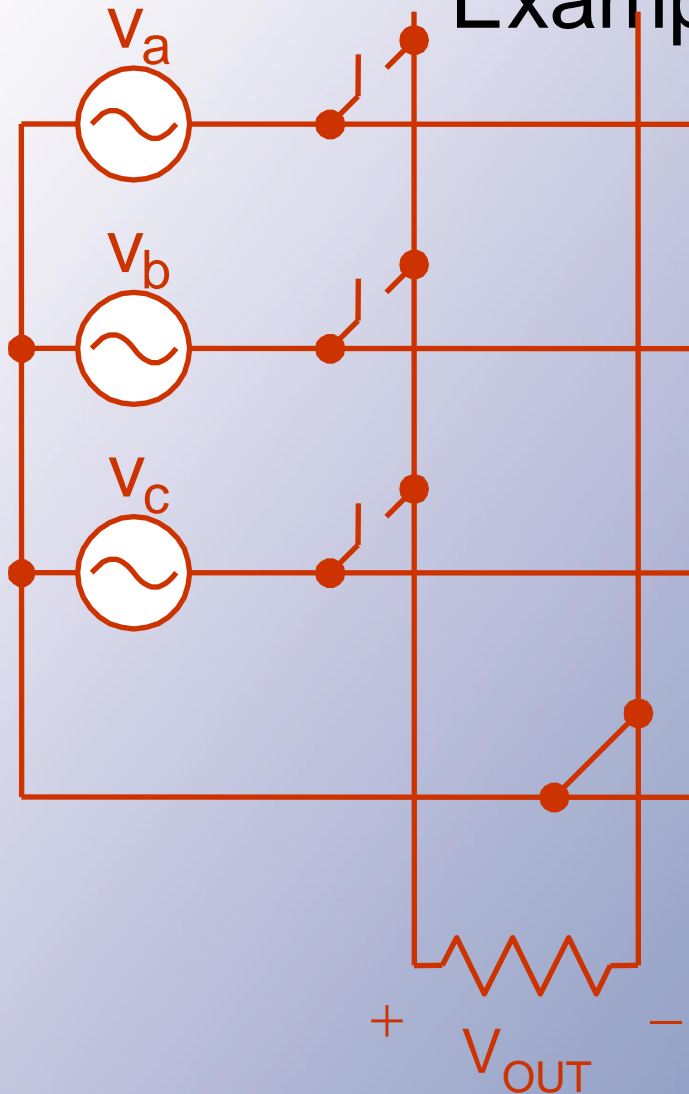
$$\text{KVL : } \sum_{i=1}^4 q_{i1} \leq 1 \qquad \sum_{i=1}^4 q_{i2} \leq 1$$

KCL : No restriction.

## Switching Functions

- We can define our *software* problem in terms of choices of switching functions.
- We can find out the expected waveforms in many types of converters.

## Example 3



$$q_{12} = 0$$

$$q_{22} = 0$$

$$q_{32} = 0$$

$$q_{41} = 0$$

$$q_{42} = 1$$

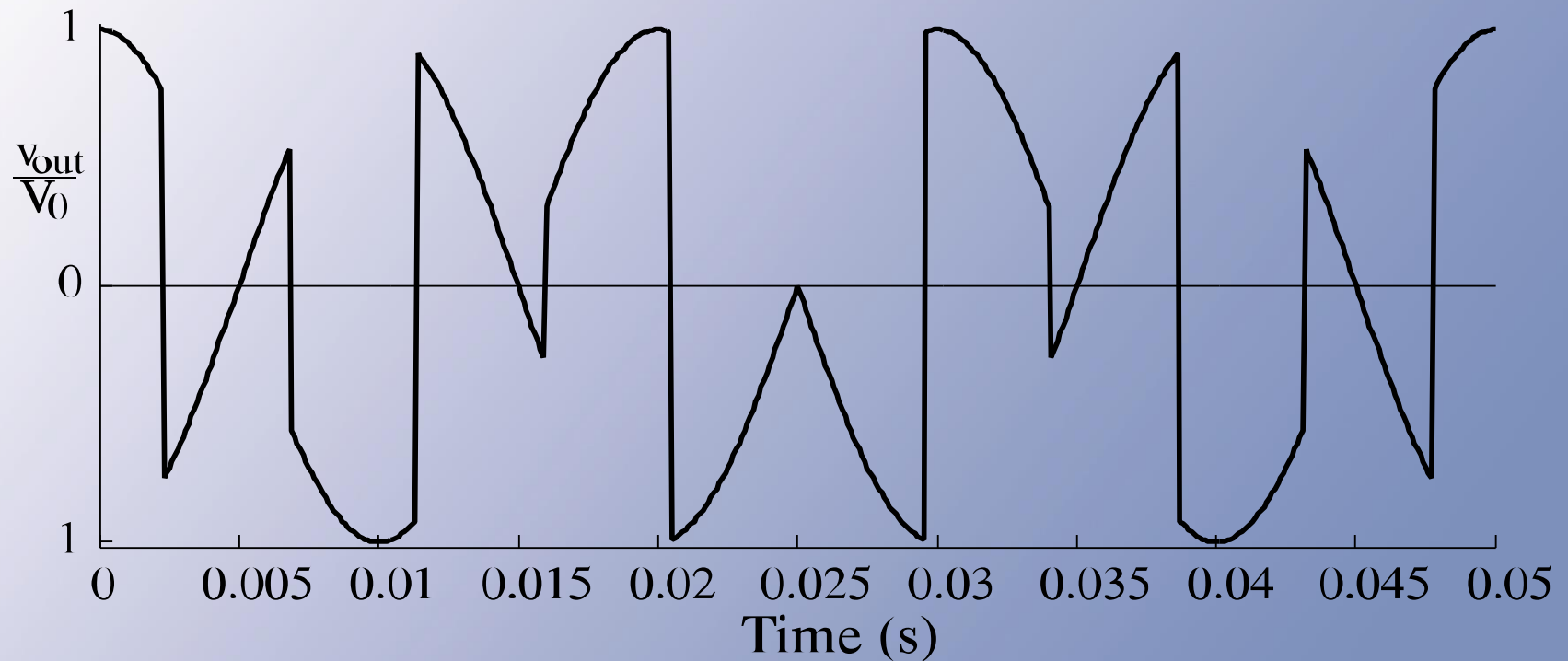
## Example 3

$$v_{out} = q_{11}v_a + q_{21}v_b + q_{31}v_c + q_{41}0 \\ - q_{12}v_a - q_{22}v_b - q_{32}v_c - q_{42}0$$

$$v_{out} = \sum_{i=1}^4 q_{i1}v_i - \sum_{i=1}^4 q_{i2}v_i$$

## Example 3

The result is a “piecewise sinusoid.”





So far

1) KVL + KCL:

Shorthand notation to understand the restrictions.

2) Outputs: (Voltage to current)

Given by products of switching functions and waveforms.

$$V_{out} = QV_{in}$$

$$I_{in} = QI_{out}$$

## Switching Devices

- Characteristics of an **ideal** switch:
  - Any polarity of  $v$  or  $i$ , and no limits.
  - Can turn on or off at any time.
  - On,  $v = 0$ . Off,  $i = 0$ .
  - Acts instantly.

## Switching Devices

- Real devices do not do any of this, of course.
- But even the best possible parts still have polarity limitations.

## Switching Devices

- Some typical devices and capabilities are given below.
- For silicon PN devices, the typical forward drop is 1 V (*not* 0.7 V).

## Diodes

- Current ratings exist from 1 A to nearly 10,000 A.
- Voltage ratings are from 10 V to 20kV.
- Not both at once (highest is about 5,000 A, 5,000 V).

## Diodes

- Power junction devices with speeds of 20 ns to about 100  $\mu$ s are used. As a general rule, large devices are slower.
- However, certain device grades are very slow.

## Diodes

- Example: 1N4004 rectifier diode, 1 A, 400 V, speed is about 2  $\mu$ s.
- MUR140 ultrafast rectifier diode, 1A, 400 V, speed is about 20 ns.

## Diodes

- Schottky diodes are also widely used in power electronics.
  - + **Lower forward drop**
  - + **Very fast**
  - Lower voltage ratings
  - Higher leakage
- Emerging: SiC Schottkys, rated to 600 V or more (e.g. 600 V, 10 A, extremely fast)



## Bipolar Transistors

- Rarely used as power switches now.
- IGBTs have replaced them in nearly all applications.

## BJT

- Speed depends on absolute and relative rating.
- Typical devices (several amps and above) switch in 500 ns to a few tens of microseconds.

## FET

- Most power devices are enhancement types that require a few volts between gate and source to turn on.
- Faster than BJTs, but lower ratings.
- Easy to use in parallel for high current.

## FET

- Individual devices to about 100 A and 1000 V.
- Maximum power (single device) is about 10 kW.
- Modular packages with multiple devices can reach 500 A.

## FET

- Power FETs are constructed as millions of small devices in parallel.
- The process inherently adds a “reverse parallel” diode internally.

## SCRs

- The generic term is *thyristor*.
- These are PNPN multi-junction devices.
- “Latching” behavior: either off, or on, with a gate pulse.
- The SCR acts like a diode when on.

## SCRs

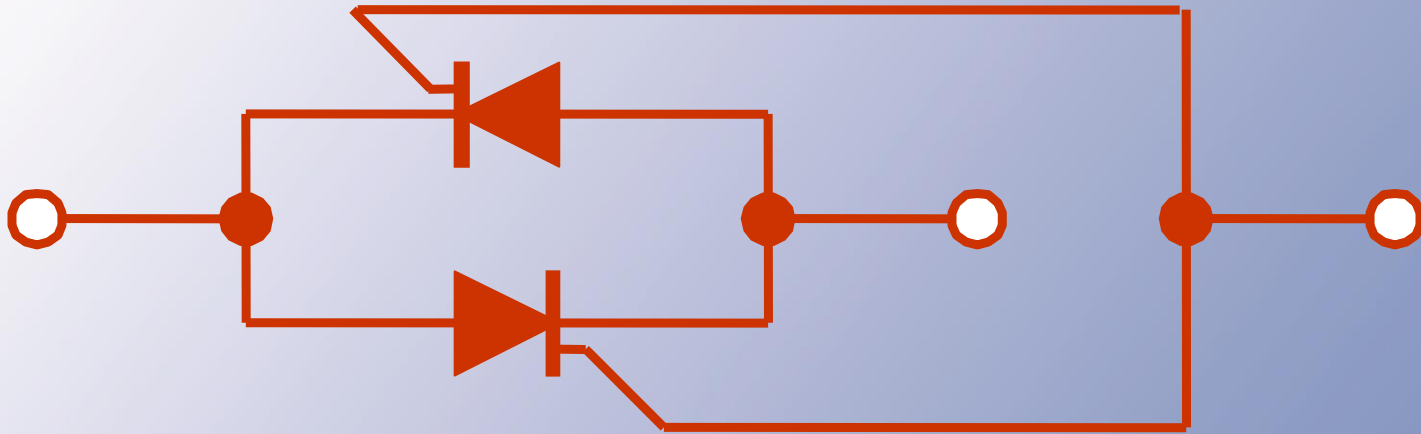
- Ratings similar to diodes. Devices that can handle 6000 A and 6000 V simultaneously are available.
- Constructed as single-wafer devices.
- Relatively slow, 1  $\mu\text{s}$  at best.

## Other Thyristors

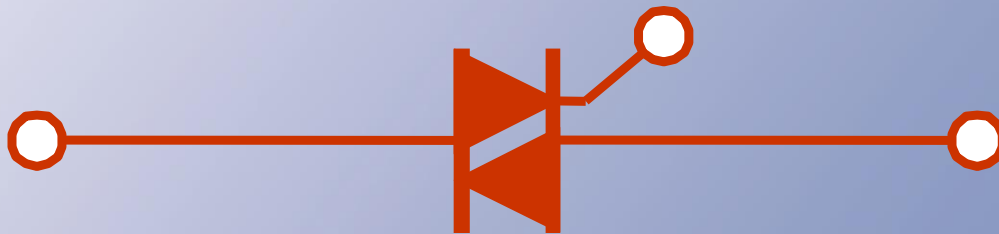
- GTO: an SCR that can be forced off with a negative gate pulse
- Light-fired SCR: an SCR that can be triggered with photons (from a laser)
- TRIAC: two SCRs in reverse parallel



## Other Thyristors



## TRIAC

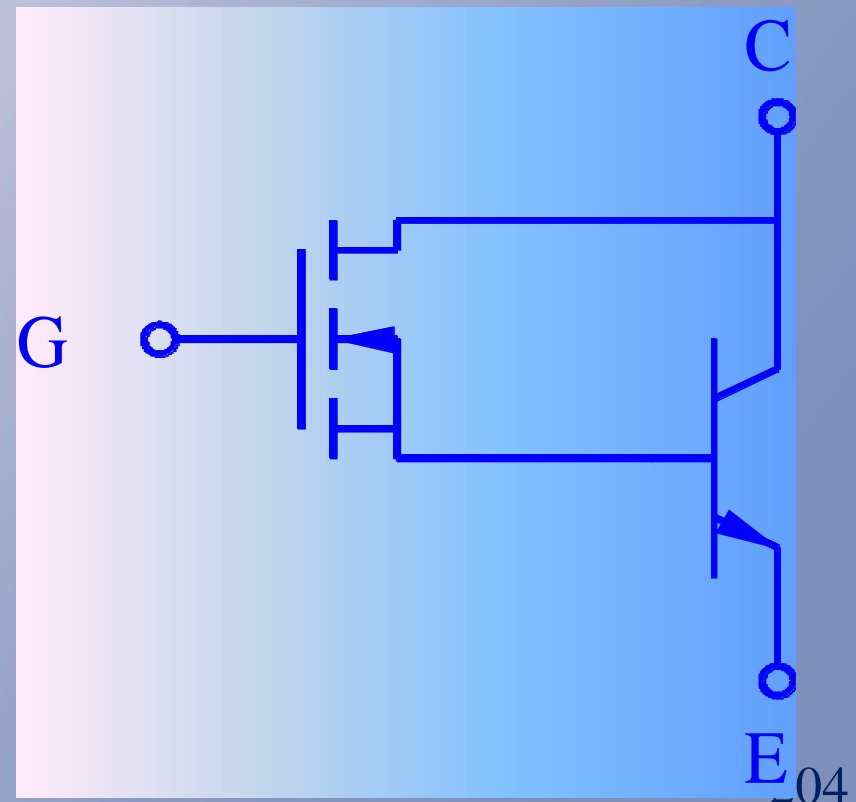
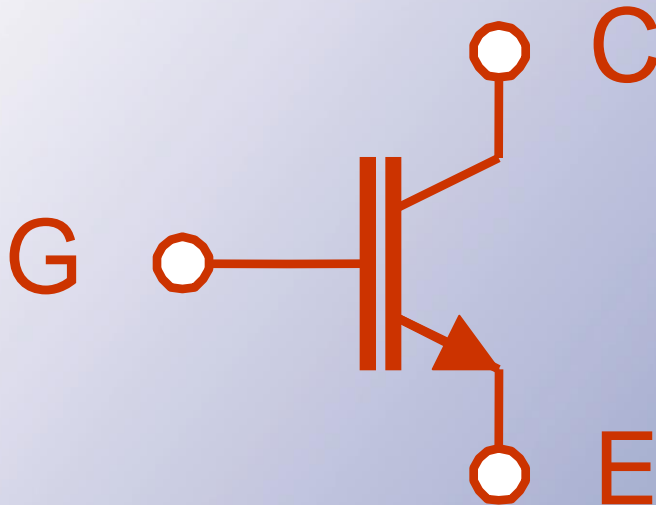


## Combined Devices

- Combination devices are becoming popular.
- The oldest is the Darlington pair of BJTs.
- The IGBT is similar to a Darlington FET/BJT combination.

## Combined Devices

- IGBT (Insulated-Gate Bipolar Transistor).
- Darlington combination of an FET and a BJT.

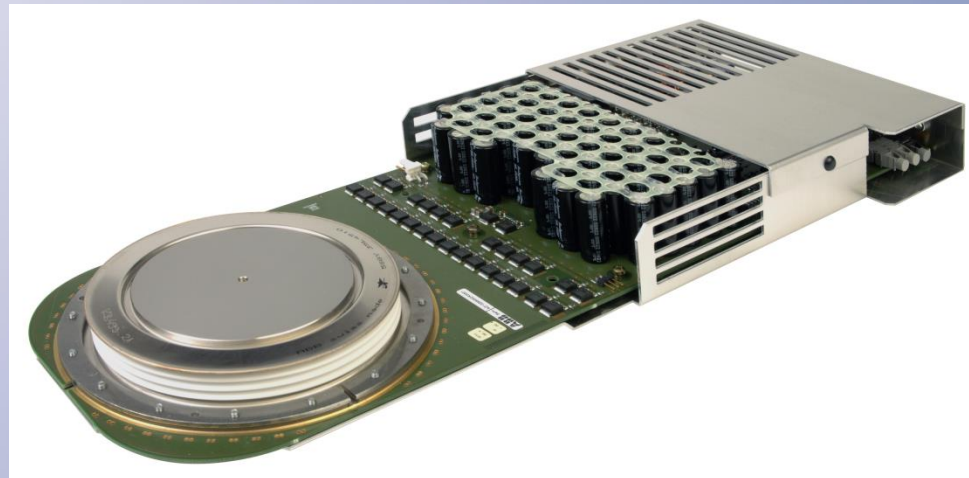


## Combined Devices

- The IGBT combines gate behavior of FET with low voltage drop of BJT.
- Very popular for inverters.
- Devices are rated up to 1200 V and 1200 A.

## Combined Devices

- IGBT is somewhat faster than BJT.
- High-power combinations are also available.
- Example: IGCT.



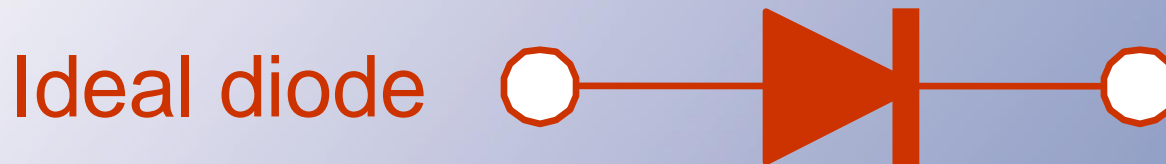


## Restricted Switches

- Semiconductors (even the best ones) have polarity limitations.
- The *restricted switch* concept represents polarity effects in an ideal way.
- Classic example: Ideal diode conducts forward, blocks reverse.

## Restricted Switch

- Idealized device **with polarity**



- $V_{\text{forward}} = 0$     $I_{\text{leak}} = 0$

## Restricted Switch Types

- **Ideal diode**
  - No forward voltage drop
  - No leakage current
  - Action is determined by terminal conditions
  - Symbol: triangle and bar
  - *Forward conducting, reverse blocking switch (FCRB)*





## Restricted Switch Types



Triangle shows carrying direction



Bar shows blocking action



Conducts in both directions. Does not block. Piece of wire



Prevents flow in both directions. Open circuit



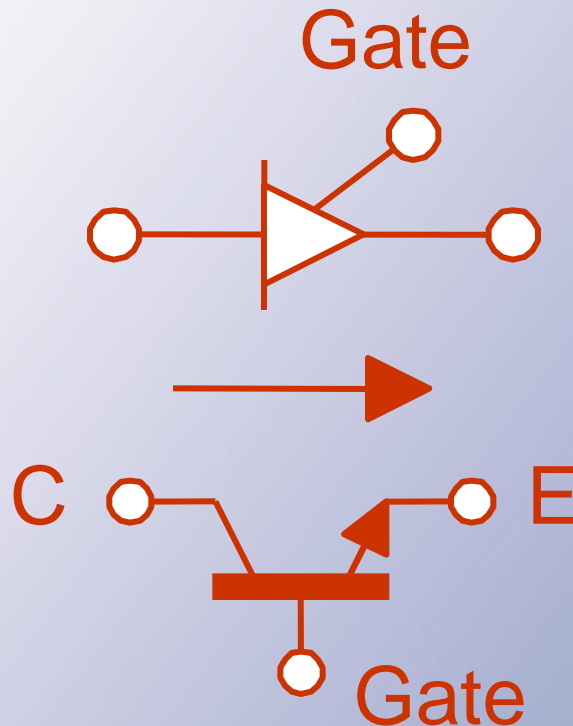
Example: **Rectifier diode**

## FCFB

- Forward-conducting forward blocking (FCFB) switch
- Conducts or blocks in forward direction
- Needs a gate to establish operation
- Action is not allowed in reverse
- Describes a BJT or IGBT



## FCFB

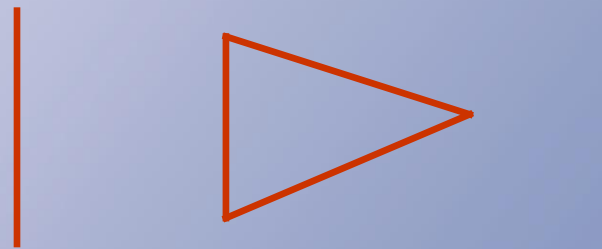


Symbol. Action not defined in reverse.

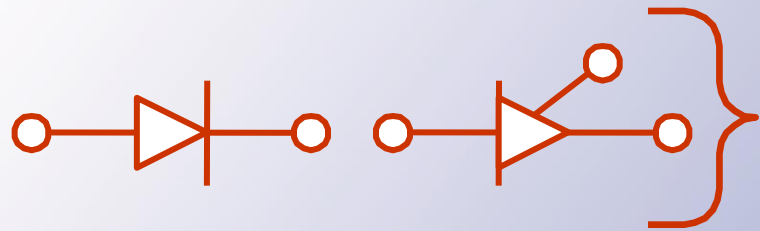
Example of implementation

## FCBB

- Forward conducting, bidirectional blocking (**FCBB**)
- Always blocks in reverse, can carry or block forward.
- This describes a **GTO**, or a **reverse-blocking IGBT**.



## FCBB

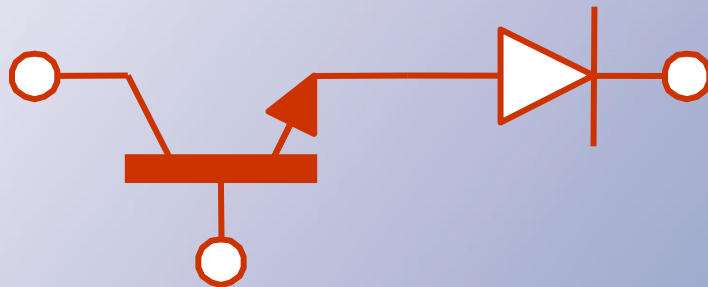


Unipolar switching devices



**FCBB** – Ex: **GTO**

Similar to SCR, but without gate restrictions

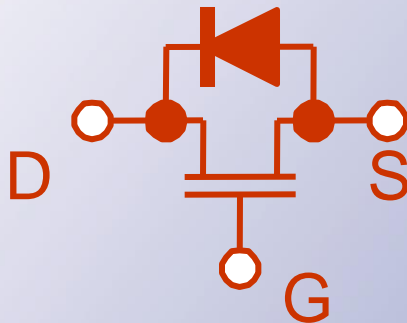
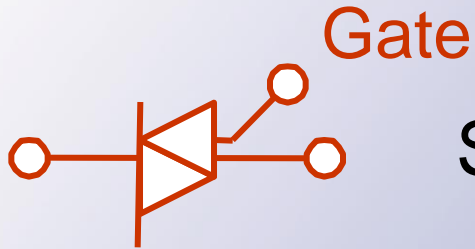


Possible implementation of FCBB

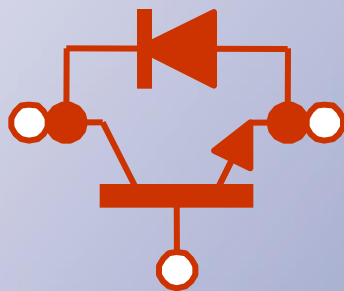
## BCFB

- Bidirectional conducting, forward blocking (**BCFB**)
- Always allows reverse flow, but can carry or block forward
- Describes an **ideal power FET**

## BCFB



Possible implementation:  
Power MOSFET



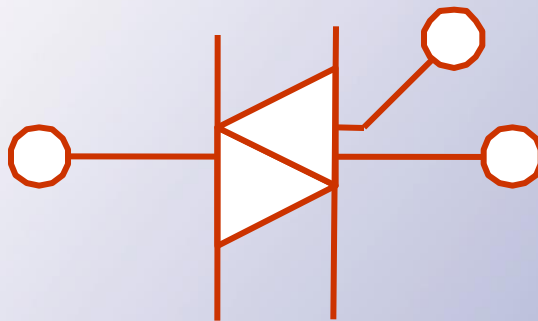
Possible implementation

## BCBB

- Bidirectional conducting, bidirectional blocking (**BCBB**)
- Describes an ideal, or **bilateral switch**  
Sometimes called a **bilat**



## BCBB

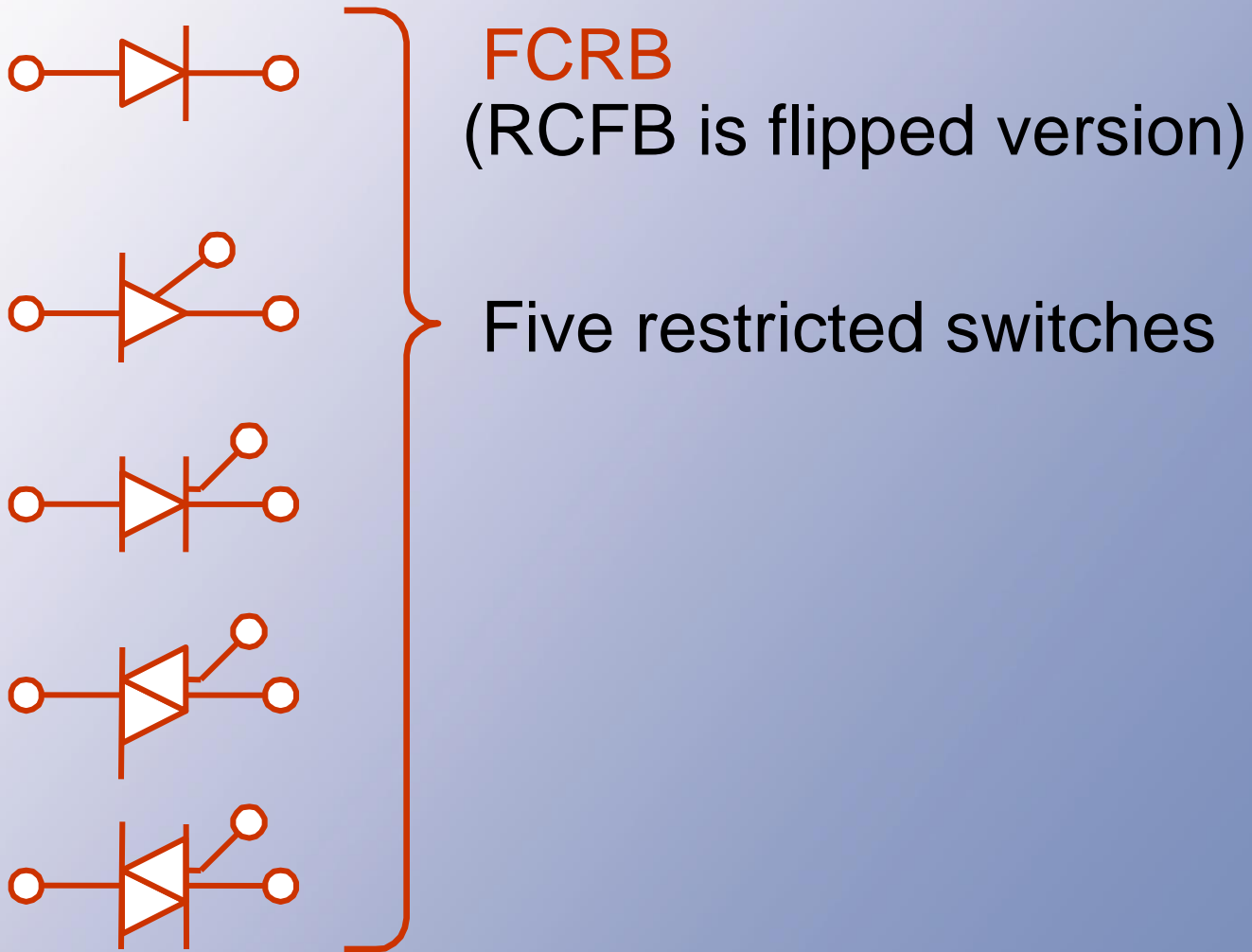


BCBB



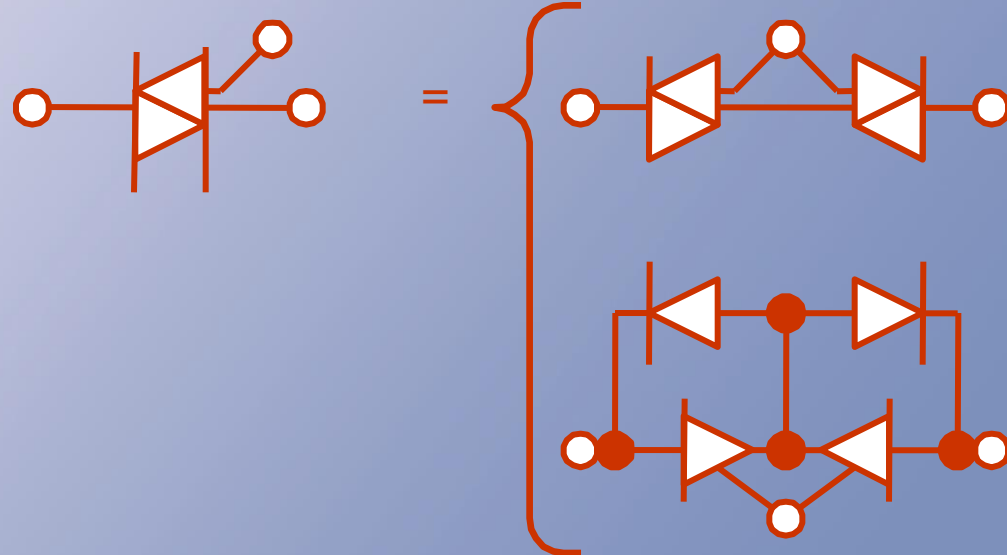
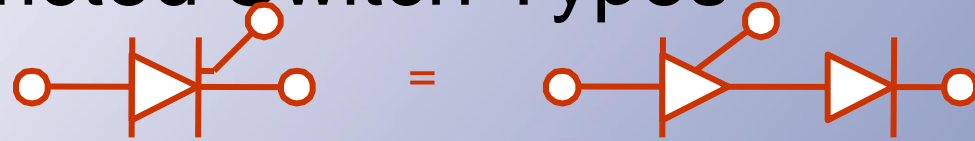
Ideal switch.

## Restricted Switch Types



## Restricted Switch Types

Making  
bilateral  
switches



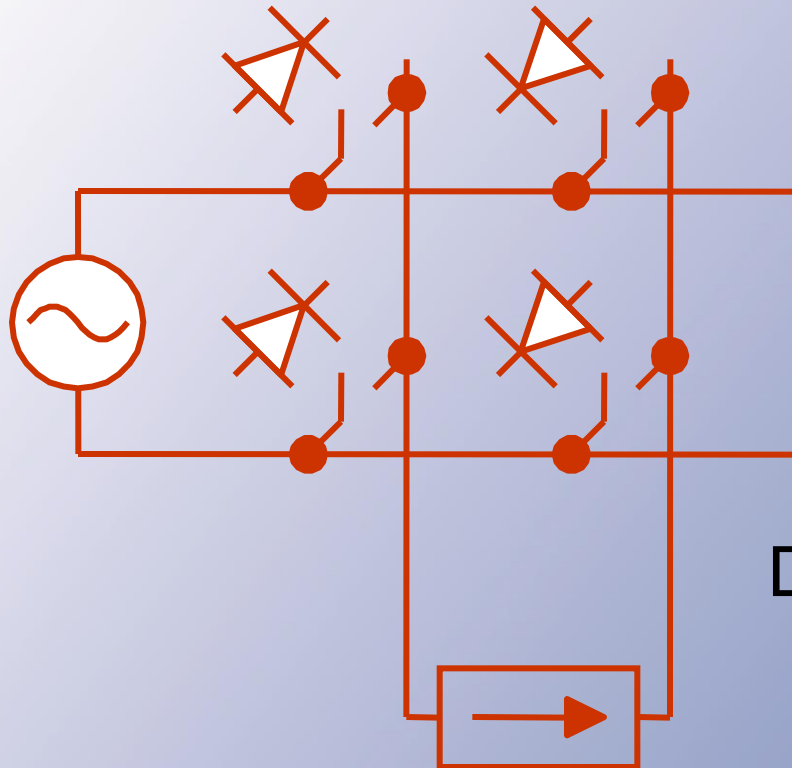
## Switch Requirements

- The specific switch requirements can be identified through a direct process:
  - Check the **current direction** when the device is **on**.
  - Check the **blocking polarity** when the device is **off**.

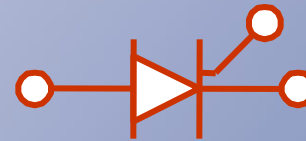
## Restricted Switch Action

- Restricted switch types can be selected based on converter function.
- Example: Rectifier (for voltage to current conversion) should block ac voltage and carry dc current.

## Restricted Switch Action AC ( $V$ ) to DC ( $I$ ) converter



AC  $V$ : Need switches to block in both directions


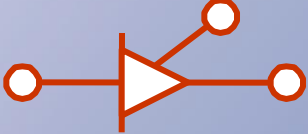
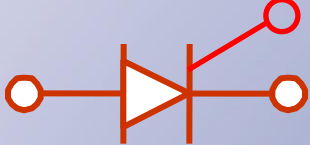
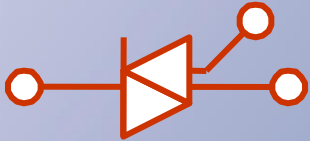
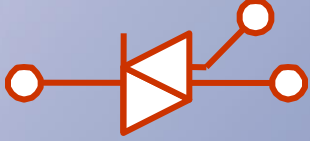
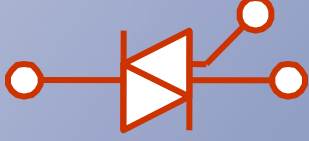


DC  $I$ : Need switches to conduct in 1 direction.

## Restricted Switch Action

- Types of converters (voltage → current)
  - Dc-dc: Unilateral devices (FCFB, FCRB)
  - Ac-dc: FCBB
  - Dc-ac: BCFB
  - Ac-ac: BCBB

## Restricted Switch Action

DC - DC Unipolar			
AC (V) - DC (I)			SCRs, GTOs
AC (I) - DC (V)			FET, BJT+Diodes, IGBT+Diodes
DC (V) - AC (I)			IGBT, FET
AC - AC			Bilateral



## Restricted Switches

- The specific switch requirements can be identified through a direct process:
  - Check the **current direction** when the device is **on**.
  - Check the **blocking polarity** when the device is **off**.



## Restricted Switches

- Choose a function (FCRB, FCFB, etc.) to match the need.
- Identify the function with a device (diode, FET, etc.).
- This is a basic approach for initial solution of the *hardware* problem.

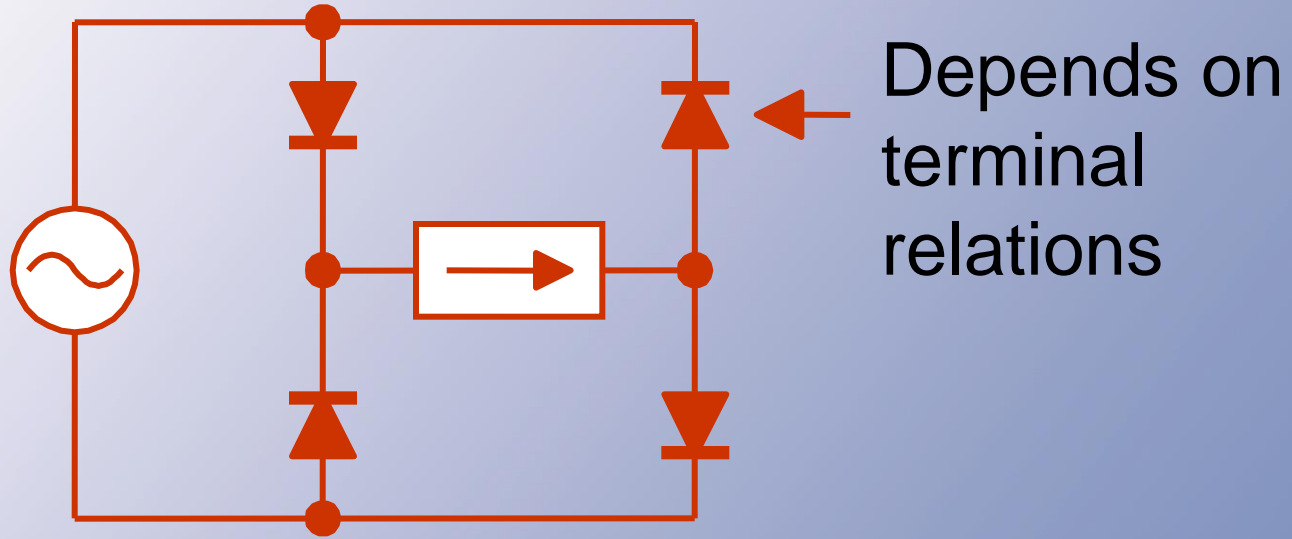
## Restricted Switches

- Restricted switches are **ideal except for polarity**. No drop, no leakage, instant action, etc.
  - FCRB = ideal diode
  - FCFB = “ideal BJT”
  - BCFB = “ideal FET”

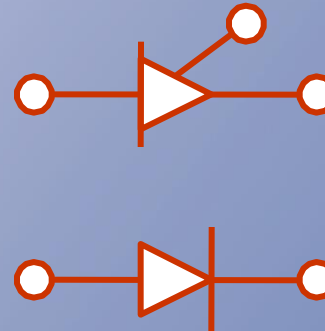
## Analysis of Diode Circuits

- The action of a diode (FCRB switch) is **determined solely by the terminal conditions**, not by an external gate.
- Once we know how to analyze diode circuits, the others follow.

## Analysis of Diode Circuits



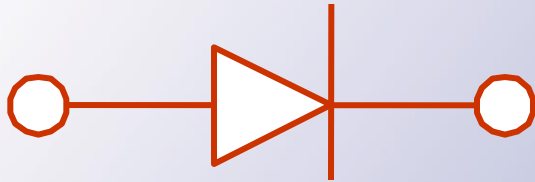
We can build any circuit with these two devices or combinations.



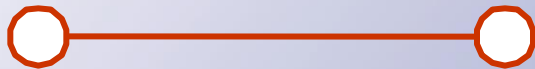
## Analysis of Diode Circuits

- The reality is that *we do not know* how to perform a direct circuit analysis when ideal diodes are present.
- *But, diodes* can only be *on or off*.
- Need to perform a “piecewise” analysis.

## Analysis of Diode Circuits



ON/OFF (which?)



Stays **ON** until  $i=0$



Stays **OFF** until  $v=0$

## Analysis of Diode Circuits

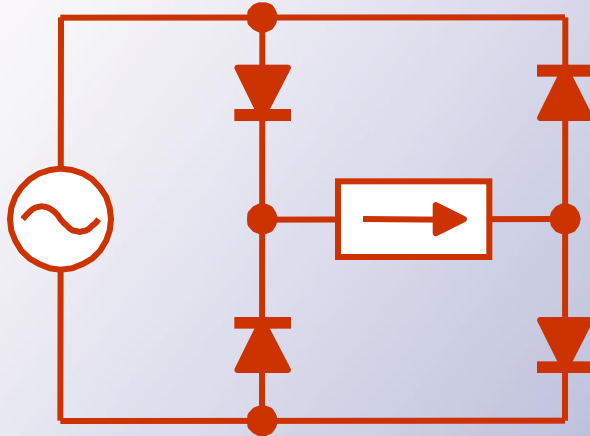
- Diodes react only to terminal conditions:
  - If the device is on, it remains on while the current is positive.
  - If the device is off, it must turn on when the voltage is positive.



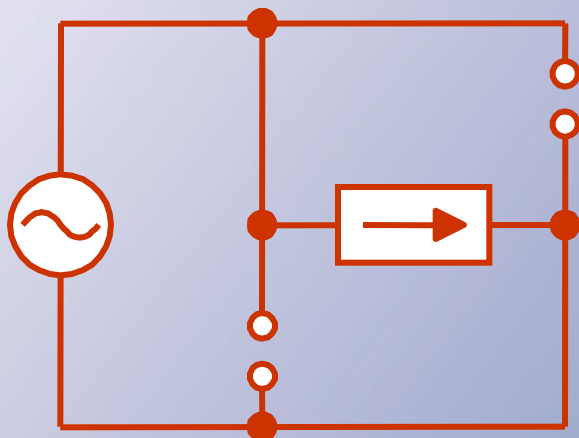
## The Trial Method

- Any diode in a circuit must be either on or off.
- The diode state (on or off) determines the circuit configuration.
- Once the configuration is known, the circuit can be drawn and analyzed.

## The Trial Method



Suppose we know which are on and which are off.



Configuration

## The Trial Method

- Although the diode states are not arbitrary, *we are free to assign states* and then check the result.
- This is a *trial and error* method. But with a little practice, there are few errors.

## The Trial Method

- Diodes satisfy KVL and KCL.
- Diodes always carry forward current, and block reverse voltage.
- In the **trial method**, we **assign diode states**, then **check for consistency**.

## The Trial Method

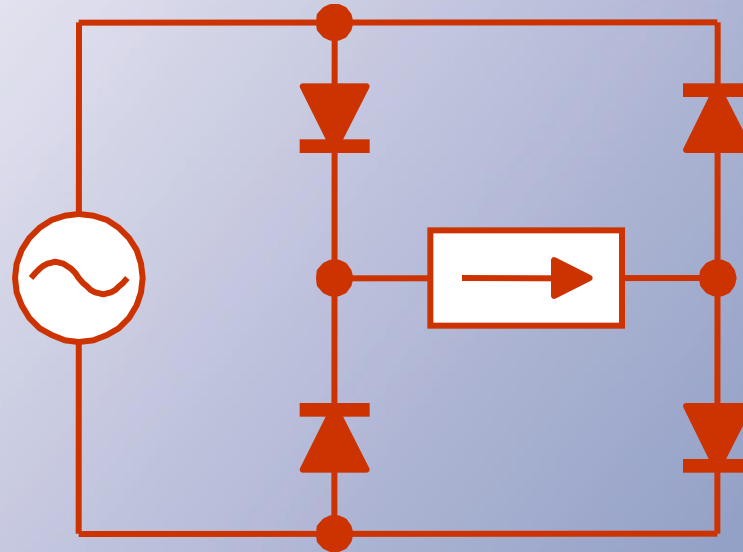
- To check:
  - KVL and KCL must be satisfied.
  - On diodes must carry  $i > 0$ .
  - Off diodes must block  $v < 0$ .
- If the checks are OK, the assigned states are valid. If not, try another.

## Keys

- Draw the configurations.
- Check polarities.
- If inconsistent, use polarities as a way to reassign states.



## Basic Example A simple rectifier connection



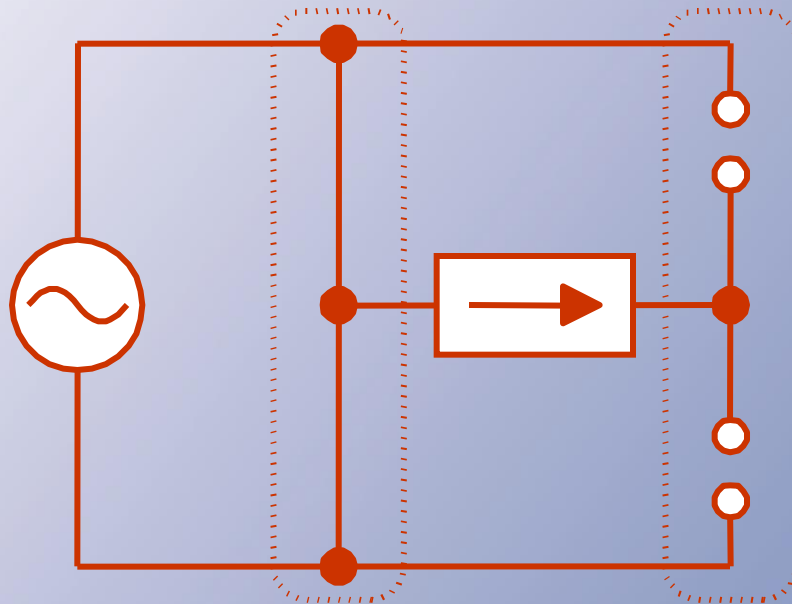
There are 16 configurations. Only one satisfies KVL, KCL, and power flow objectives.

## Basic Example

- Try a few configurations with the trial method.

They cannot be  
**on** at the  
same time.

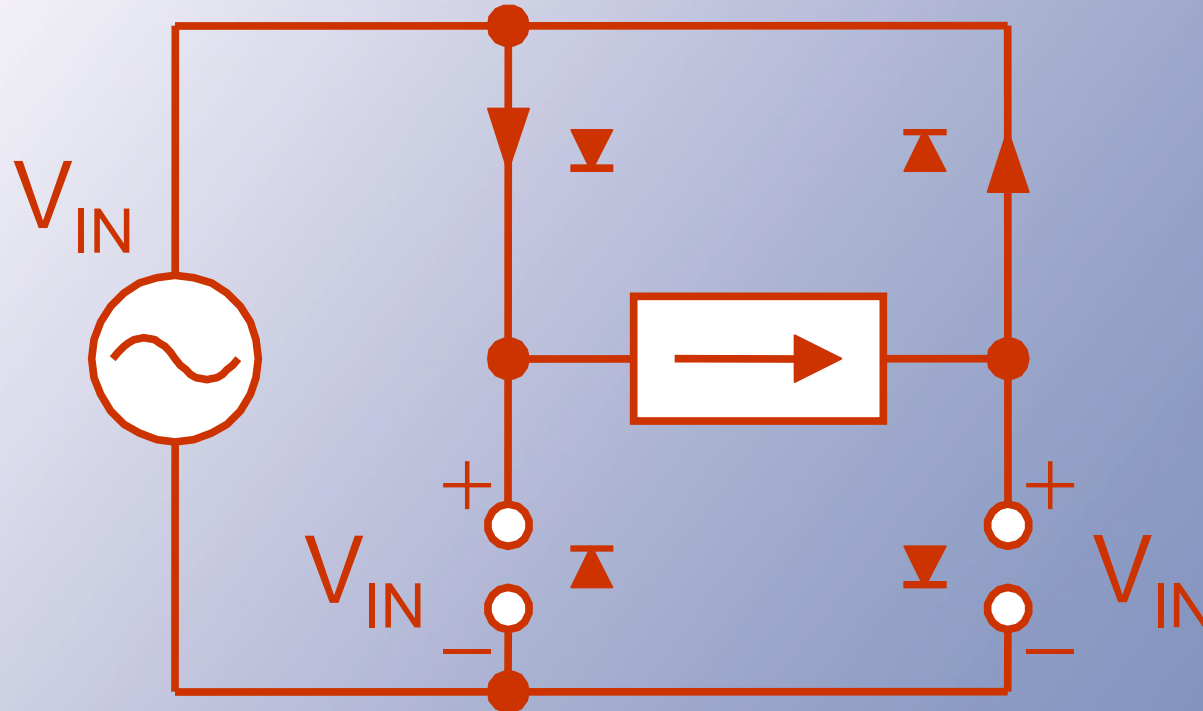
They cannot be  
**off** at the  
same time.



KVL and KCL  
not consistent

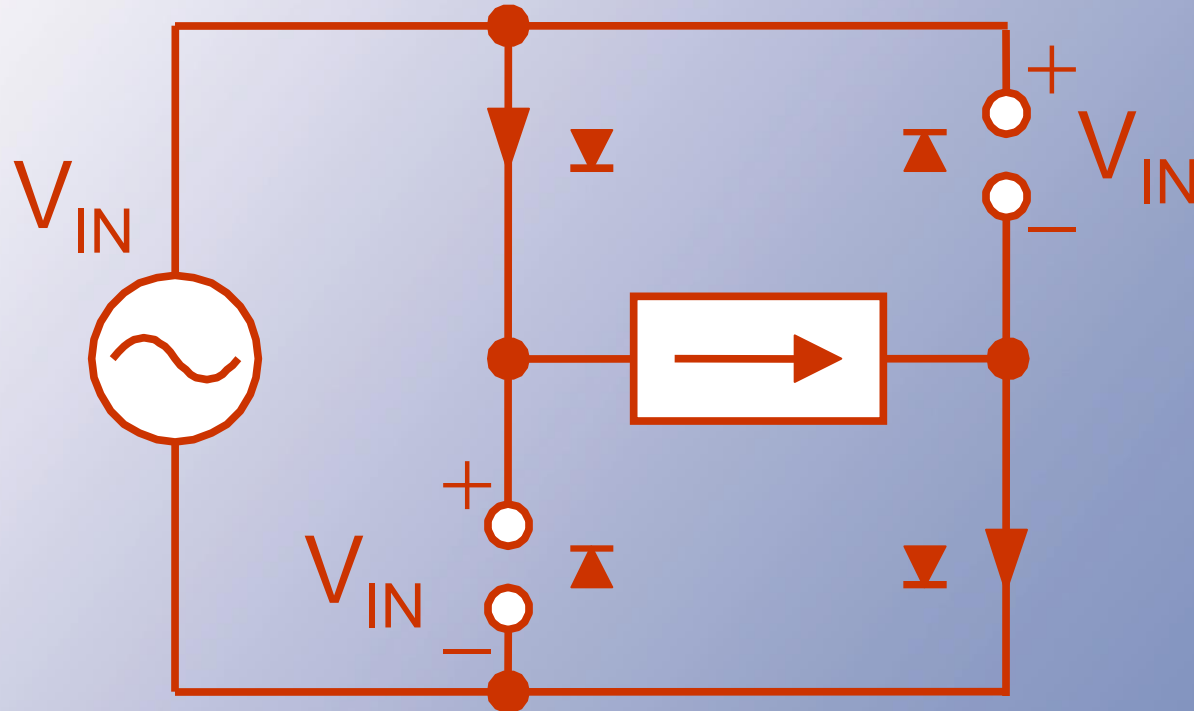


## Basic Example



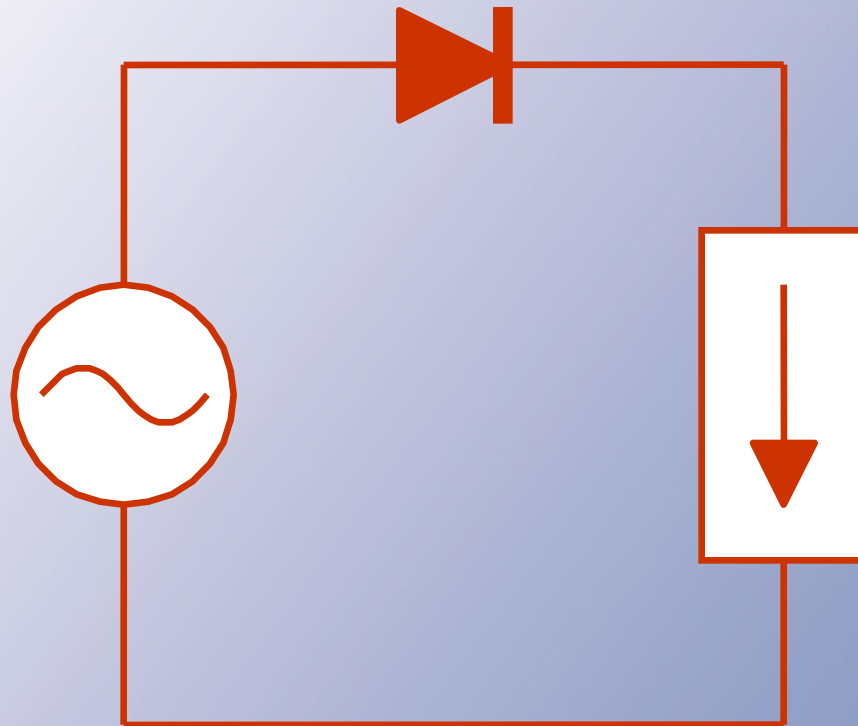
- KVL, KCL: Ok
- ON-State: Ok OFF-State: Inconsistent.

## Basic Example



- KVL, KCL: Ok
- ON-State: Ok OFF-Voltage: Consistent if  $V_{in} > 0$

## Basic Example



Diode is always on. Voltage is not relevant.  
Current requires ON state.

## The Trial Method

- The method works for complex circuits.
- **Diodes** are **passive** switches, since their action is governed by terminal conditions.
- Switches with **gates** are **active**.

## The Trial Method

- Devices such as the **FET** can be analyzed as **combinations**:
  - If the gate is high, the device is on.
  - If the gate is low, the reverse parallel diode must be checked.
- The FCBB switch is similar (dual).

## Summary

- Restricted switches, combined with the trial method, allow us to analyze idealized power converters.
- New power semiconductors try to approach ideal behavior.